

Dual and Quad Micropower Rail-to-Rail Input and Output Op Amps

FEATURES

- Rail-to-Rail Input and Output
- Single Supply Input Range: -0.4V to 44V
- Micropower: 50 μ A/Amplifier Max
- MSOP Package
- Specified on 3V, 5V and \pm 15V Supplies
- High Output Current: 20mA
- Output Drives 10,000pF with Output Compensation
- Reverse Battery Protection to 18V
- No Supply Sequencing Problems
- High Voltage Gain: 1500V/mV
- High CMRR: 98dB
- No Phase Reversal
- Gain Bandwidth Product: 200kHz

APPLICATIONS

- Battery- or Solar-Powered Systems
- Portable Instrumentation
- Sensor Conditioning
- Supply Current Sensing
- Battery Monitoring
- Micropower Active Filters
- 4mA to 20mA Transmitters

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DESCRIPTION

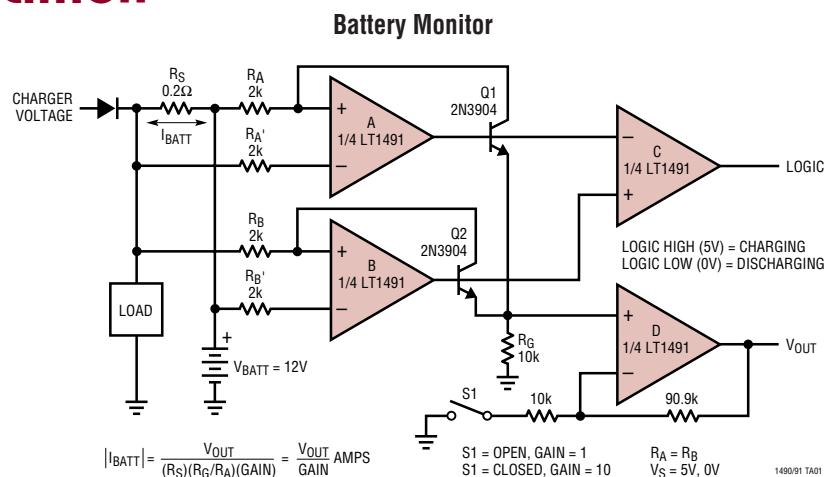
The dual LT®1490 and quad LT1491 op amps operate on all single and split supplies with a total voltage of 2V to 44V drawing only 40 μ A of quiescent current per amplifier. These amplifiers are reverse supply protected; they draw no current for reverse supply up to 18V. The input range of the LT1490/LT1491 includes both supplies and the output swings to both supplies. Unlike most micropower op amps, the LT1490/LT1491 can drive heavy loads; their rail-to-rail outputs drive 20mA. The LT1490/LT1491 are unity-gain stable and drive all capacitive loads up to 10,000pF when optional 0.22 μ F and 150 Ω compensation is used.

The LT1490/LT1491 have a unique input stage that operates and remains high impedance when above the positive supply. The inputs take 44V both differential and common mode even when operating on a 3V supply. Built-in resistors protect the inputs for faults below the negative supply up to 22V. There is no phase reversal of the output for inputs 22V below V⁻ or 44V above V⁺, independent of V⁺.

The LT1490 dual op amp is available in the 8-pin MSOP, SO and PDIP packages. The quad LT1491 is available in the 14-pin SO and PDIP packages.

For new designs it is recommended that the LT1490A/LT1491A be used instead of the LT1490/LT1491. See the LT1490A/LT1491A data sheet for details.

TYPICAL APPLICATION



ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-)	44V	Junction Temperature	150°C
Input Differential Voltage	44V	Specified Temperature Range (Note 3) ..	-40°C to 85°C
Input Current	$\pm 25\text{mA}$	Storage Temperature Range	-65°C to 150°C
Output Short-Circuit Duration (Note 2)	Continuous	Lead Temperature (Soldering, 10 sec)	300°C
Operating Temperature Range	-40°C to 85°C		

PACKAGE/ORDER INFORMATION (For new designs, use the LT1490A or LT1491A)

TOP VIEW	ORDER PART NUMBER	TOP VIEW		ORDER PART NUMBER
	LT1490CMS8			LT1491CN
MS8 PACKAGE 8-LEAD MSOP	LT1490CN8	N8 PACKAGE 8-LEAD PDIP	LT1491CS	LT1491CS
S8 PACKAGE 8-LEAD PLASTIC SO	LT1490CS8		LT1491IN	LT1491IN
$T_{JMAX} = 150^\circ\text{C}, \theta_{JA} = 250^\circ\text{C/W}$ (MS8)	LT1490IN8			LT1491IS
$T_{JMAX} = 150^\circ\text{C}, \theta_{JA} = 130^\circ\text{C/W}$ (N8)	LT1490IS8			
$T_{JMAX} = 150^\circ\text{C}, \theta_{JA} = 190^\circ\text{C/W}$ (S8)				
MS8 PART MARKING		TOP VIEW		MS8 PART MARKING
LTBB		TOP VIEW		S8 PART MARKING
S8 PART MARKING		TOP VIEW		1490 1490I
		N PACKAGE 14-LEAD PDIP S PACKAGE $T_{JMAX} = 150^\circ\text{C}, \theta_{JA} = 110^\circ\text{C/W}$ (N)		
		$T_{JMAX} = 150^\circ\text{C}, \theta_{JA} = 150^\circ\text{C/W}$ (S)		

Consult factory for Military grade parts.

ELECTRICAL CHARACTERISTICSThe ● denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 3\text{V}, 0\text{V}$; $V_S = 5\text{V}, 0\text{V}$; $V_{CM} = V_{OUT} = \text{half supply}$, unless otherwise noted. (Note 3)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT1490 N Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$		220	800	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		1000	μV
		LT1490 S Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$		220	950	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		1200	μV
	LT1491 N Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$		300	1100		μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		1350	μV
		LT1490CMS8 Package, LT1491 S Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$		350	1450	μV
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●		1650	μV
$V_{OS\ TC}$	Input Offset Voltage Drift	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ (Note 7)	●	2	4	$\mu\text{V}/^\circ\text{C}$
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$ (Note 7)	●	2	4	$\mu\text{V}/^\circ\text{C}$
I_{OS}	Input Offset Current	$V_{CM} = 44\text{V}$ (Note 4)	●	0.2	0.8	nA
I_B	Input Bias Current	$V_{CM} = 44\text{V}$ (Note 4)	●	4	8	nA
		$V_S = 0\text{V}$	●	4	10	μA
				0.1	0.1	nA
	Input Noise Voltage	0.1Hz to 10Hz		1		$\mu\text{V}_{\text{P-P}}$

ELECTRICAL CHARACTERISTICS

The ● denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 3\text{V}, 0\text{V}$; $V_S = 5\text{V}, 0\text{V}$; $V_{CM} = V_{OUT} = \text{half supply}$, unless otherwise noted. (Note 3)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
e_n	Input Noise Voltage Density	$f = 1\text{kHz}$		50		$\text{nV}/\sqrt{\text{Hz}}$
i_n	Input Noise Current Density	$f = 1\text{kHz}$		0.03		$\text{pA}/\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	Differential Common Mode, $V_{CM} = 0\text{V}$ to 44V	6 4	17 11		$\text{M}\Omega$ $\text{M}\Omega$
C_{IN}	Input Capacitance			4.6		pF
	Input Voltage Range		●	0	44	V
CMRR	Common Mode Rejection Ratio (Note 4)	$V_{CM} = 0\text{V}$ to $V_{CC} - 1\text{V}$	●	84	98	dB
		$V_{CM} = 0\text{V}$ to 44V	●	80	98	dB
A_{VOL}	Large-Signal Voltage Gain	$V_S = 3\text{V}$, $V_0 = 500\text{mV}$ to 2.5V , $R_L = 10\text{k}$ $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		200 133 100	1500	V/mV V/mV V/mV
		$V_S = 5\text{V}$, $V_0 = 500\text{mV}$ to 4.5V , $R_L = 10\text{k}$ $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		400 250 200	1500	V/mV V/mV V/mV
		$V_S = 3\text{V}$, No Load	●	22	50	mV
		$V_S = 3\text{V}$, $I_{SINK} = 5\text{mA}$	●	250	450	mV
V_{OL}	Output Voltage Swing Low	$V_S = 5\text{V}$, No Load	●	22	50	mV
		$V_S = 5\text{V}$, $I_{SINK} = 5\text{mA}$	●	250	500	mV
		$V_S = 5\text{V}$, $I_{SINK} = 10\text{mA}$		330	500	mV
		$V_S = 3\text{V}$, No Load	●	2.95	2.978	V
V_{OH}	Output Voltage Swing High	$V_S = 3\text{V}$, $I_{SOURCE} = 5\text{mA}$	●	2.55	2.6	V
		$V_S = 5\text{V}$, No Load	●	4.95	4.978	V
		$V_S = 5\text{V}$, $I_{SOURCE} = 10\text{mA}$	●	4.30	4.6	V
		$V_S = 3\text{V}$, Short to GND		10	15	mA
I_{SC}	Short-Circuit Current (Note 2)	$V_S = 3\text{V}$, Short to V_{CC}		10	30	mA
		$V_S = 5\text{V}$, Short to GND		15	25	mA
		$V_S = 5\text{V}$, Short to V_{CC}		15	30	mA
		$V_S = 2.5\text{V}$ to 12.5V , $V_{CM} = V_0 = 1\text{V}$	●	84	98	dB
PSRR	Power Supply Rejection Ratio		●	2	2.5	V
		Minimum Operating Supply Voltage	●			
I_S	Supply Current per Amplifier (Note 5)	Reverse Supply Voltage	●	18	27	V
		$I_S = -100\mu\text{A}$ per Amplifier	●		40 50 55	μA
GBW	Gain Bandwidth Product (Note 4)	$f = 1\text{kHz}$		110	180	kHz
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	100		kHz
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●	90		kHz
SR	Slew Rate (Note 6)	$A_V = -1$, $R_L = \infty$		0.035	0.06	$\text{V}/\mu\text{s}$
		$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$	●	0.031		$\text{V}/\mu\text{s}$
		$-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	●	0.030		$\text{V}/\mu\text{s}$

ELECTRICAL CHARACTERISTICS

The ● denotes specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$, $V_{OUT} = 0\text{V}$, unless otherwise noted. (Note 3)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage	LT1490 N, S Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		250	1200	μV
			●	1400		μV
			●	1500		μV
	LT1491 N Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$			350	1250	μV
			●	1500		μV
			●	1600		μV
	LT1490CMS8 Package, LT1491 S Package $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$			400	1600	μV
			●	1850		μV
			●	1950		μV
$V_{OS\ TC}$	Input Offset Voltage Drift	$0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ (Note 7) $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$ (Note 7)	●	3	6	$\mu\text{V}/^\circ\text{C}$
●	Input Offset Current		●	3	6	$\mu\text{V}/^\circ\text{C}$
I_B	Input Bias Current		●	0.2	0.8	nA
	Input Noise Voltage	0.1Hz to 10Hz		4	8	nA
e_n	Input Noise Voltage Density	f = 1kHz		1		$\mu\text{V}_{\text{P-P}}$
i_n	Input Noise Current Density	f = 1kHz		50		$\text{nV}/\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	Differential Common Mode, $V_{CM} = -15\text{V}$ to 14V		6	17	$\text{M}\Omega$
				15000		$\text{M}\Omega$
C_{IN}	Input Capacitance			4.6		pF
	Input Voltage Range		●	-15	29	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -15\text{V}$ to 29V	●	80	98	dB
A_{VOL}	Large-Signal Voltage Gain $V_O = \pm 14\text{V}$, $R_L = 10\text{k}\Omega$ $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		100	250		V/mV
			●	75		V/mV
			●	50		V/mV
V_0	Output Voltage Swing $I_{OUT} = \pm 5\text{mA}$ $I_{OUT} = \pm 10\text{mA}$	No Load	●	± 14.9	± 14.978	V
			●	± 14.5	± 14.750	V
			●	± 14.5	± 14.670	V
I_{SC}	Short-Circuit Current (Note 2) $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$	Short to GND	●	± 20	± 25	mA
			●	± 15		mA
			●	± 10		mA
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.25\text{V}$ to $\pm 22\text{V}$	●	88	98	dB
I_S	Supply Current per Amplifier			50	70	μA
					85	μA
GBW	Gain Bandwidth Product $f = 1\text{kHz}$ $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		125	200		kHz
			●	110		kHz
			●	100		kHz
SR	Slew Rate $A_V = -1$, $R_L = \infty$, $V_0 = \pm 10\text{V}$, Measure at $V_0 = \pm 5\text{V}$ $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$ $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$		0.0375	0.07		$\text{V}/\mu\text{s}$
			●	0.0330		$\text{V}/\mu\text{s}$
			●	0.0300		$\text{V}/\mu\text{s}$

Note 1: Absolute Maximum Ratings are those values beyond which the life of the device may be impaired.

Note 2: A heat sink may be required to keep the junction temperature below absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted.

Note 3: The LT1490C/LT1491C are guaranteed to meet specified performance from 0°C to 70°C . The LT1490C/LT1491C are designed, characterized and expected to meet specified performance from -40°C to 85°C but are not tested or QA sampled at these temperatures.

The LT1490/LT1491 are guaranteed to meet specified performance from -40°C to 85°C .

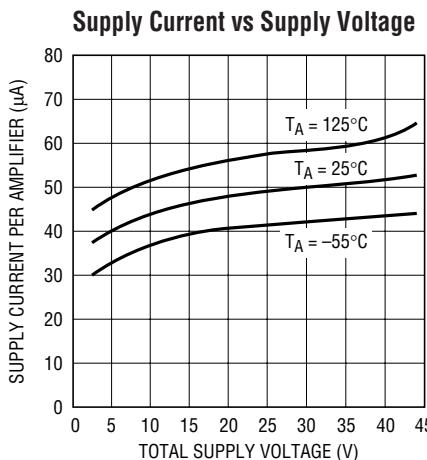
Note 4: $V_S = 5\text{V}$ limits are guaranteed by correlation to $V_S = 3\text{V}$ and $V_S = \pm 15\text{V}$ tests.

Note 5: $V_S = 3\text{V}$ limits are guaranteed by correlation to $V_S = 5\text{V}$ and $V_S = \pm 15\text{V}$ tests.

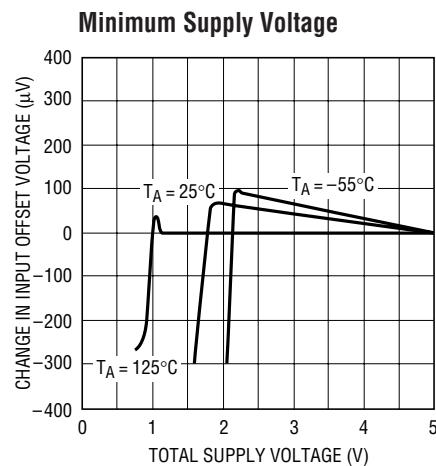
Note 6: Guaranteed by correlation to slew rate at $V_S = \pm 15\text{V}$ and GBW at $V_S = 3\text{V}$ and $V_S = \pm 15\text{V}$ tests.

Note 7: This parameter is not 100% tested.

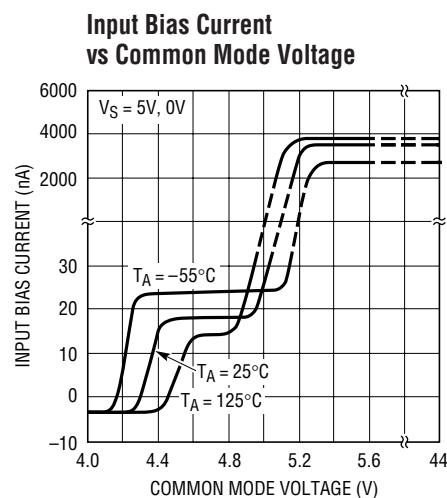
TYPICAL PERFORMANCE CHARACTERISTICS



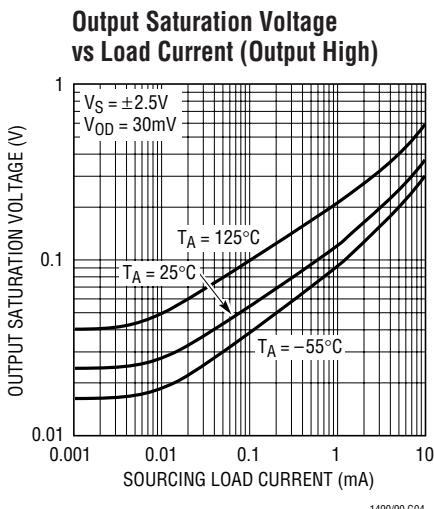
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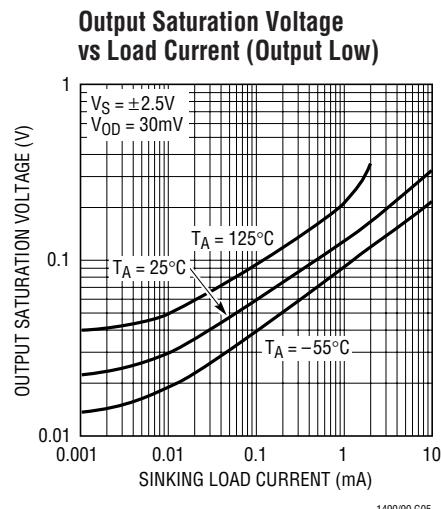
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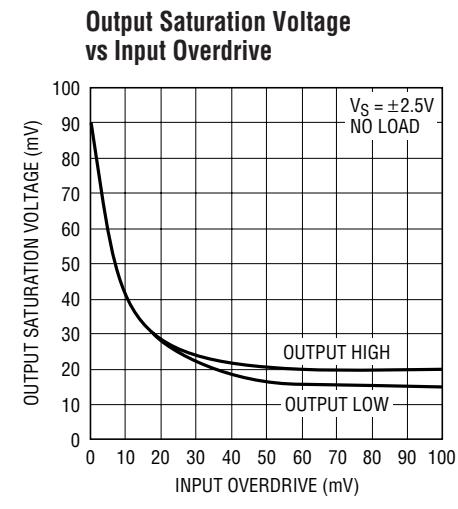
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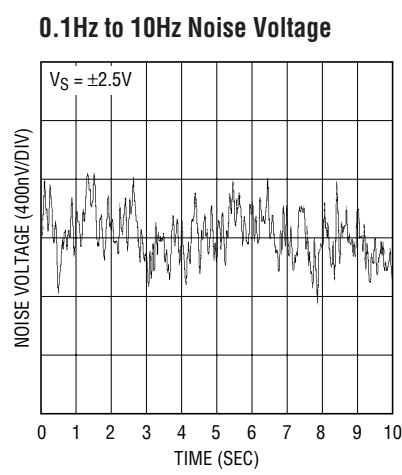
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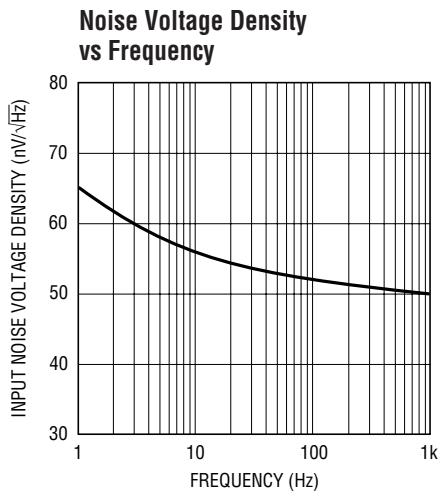
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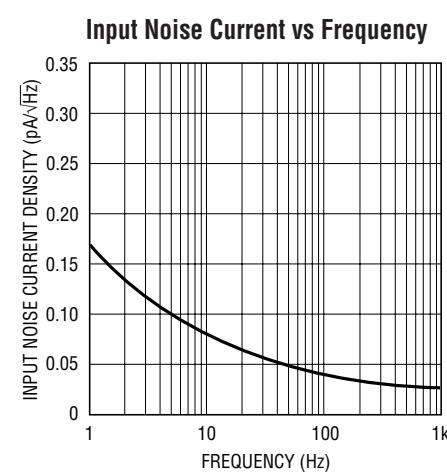
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1490 G07



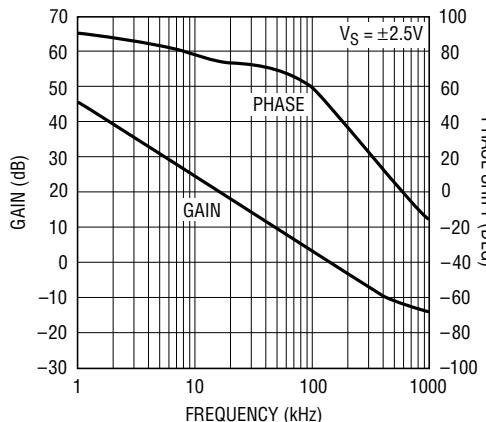
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1490/91 G09

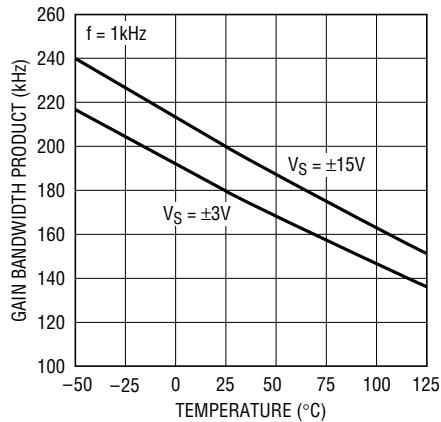
TYPICAL PERFORMANCE CHARACTERISTICS

Gain and Phase Shift vs Frequency



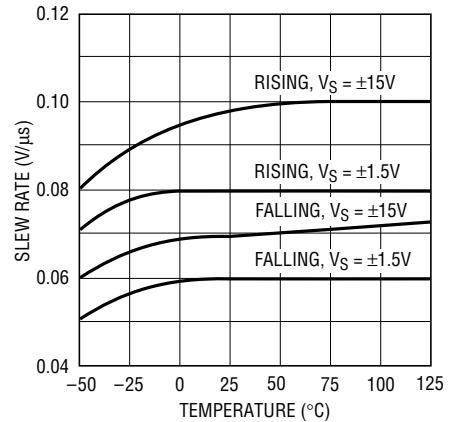
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Gain Bandwidth Product vs Temperature



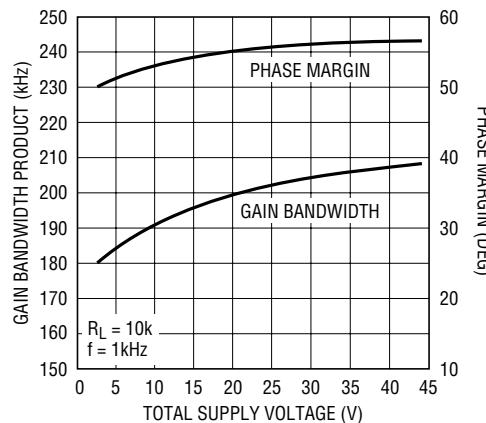
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Slew Rate vs Temperature



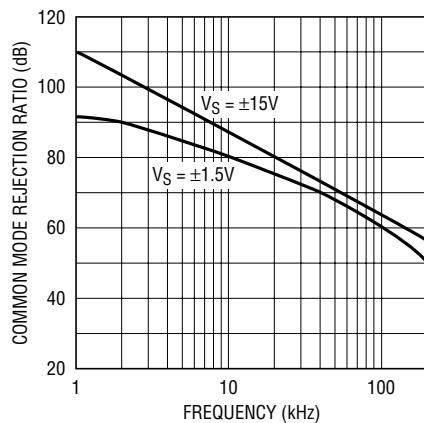
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Gain Bandwidth Product and Phase Margin vs Supply Voltage



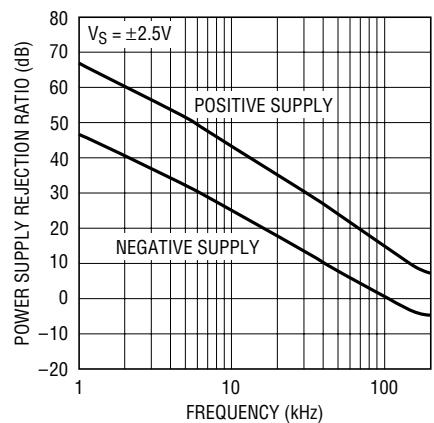
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CMRR vs Frequency



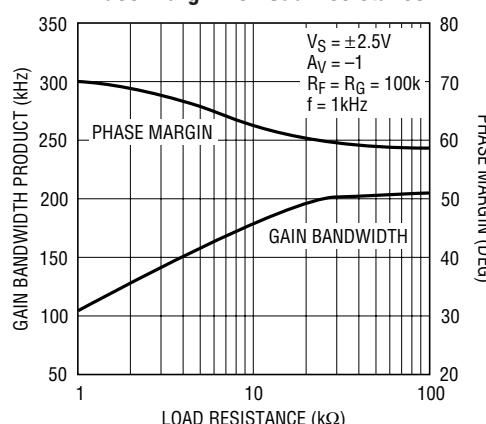
1490 G14

PSRR vs Frequency



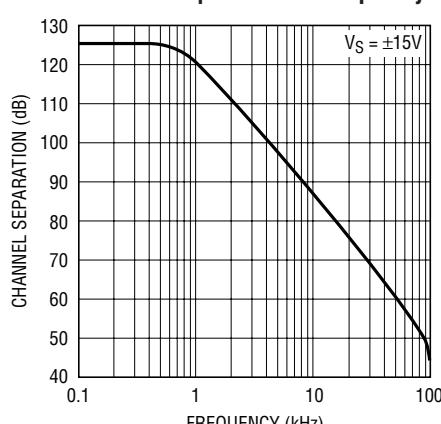
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Gain Bandwidth Product and Phase Margin vs Load Resistance



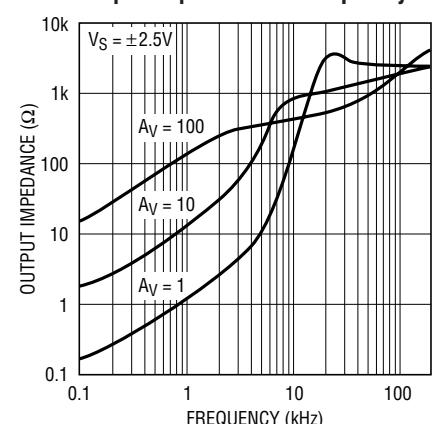
1490/91 G16

Channel Separation vs Frequency



1490/91 G17

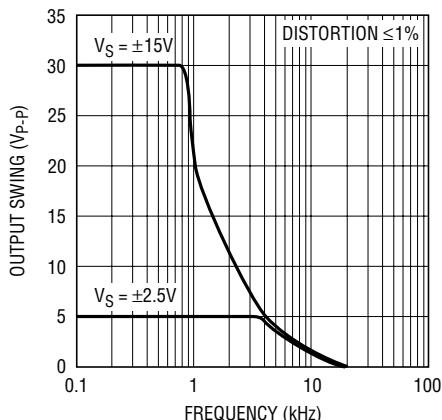
Output Impedance vs Frequency



1490/91 G18

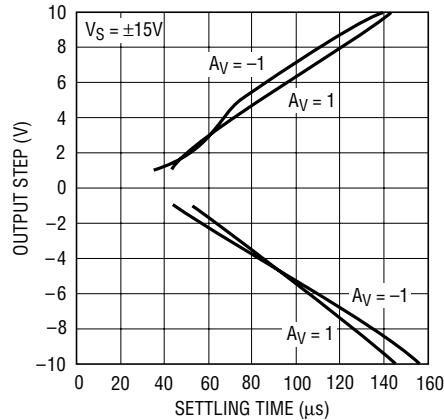
TYPICAL PERFORMANCE CHARACTERISTICS

Undistorted Output Swing vs Frequency



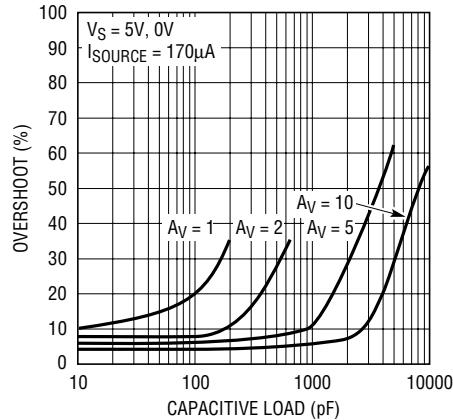
1490/91 G19

Settling Time to 0.1% vs Output Step



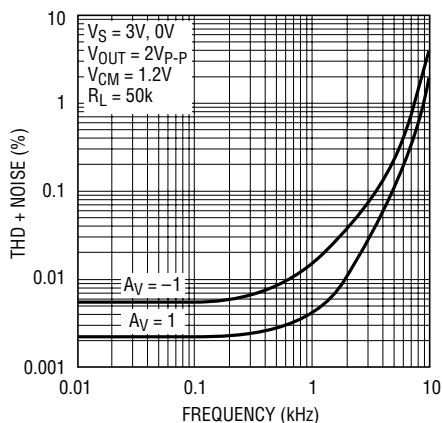
1490/91 F20

Capacitive Load Handling, Overshoot vs Capacitive Load



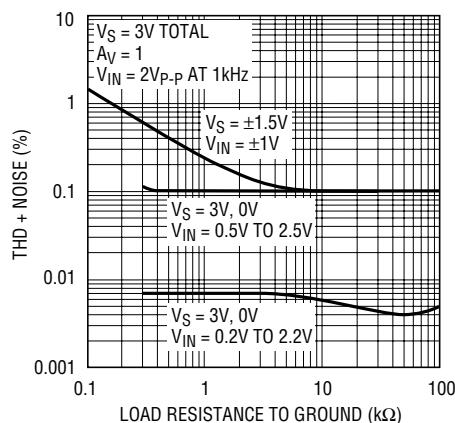
1490/91 G21

Total Harmonic Distortion + Noise vs Frequency



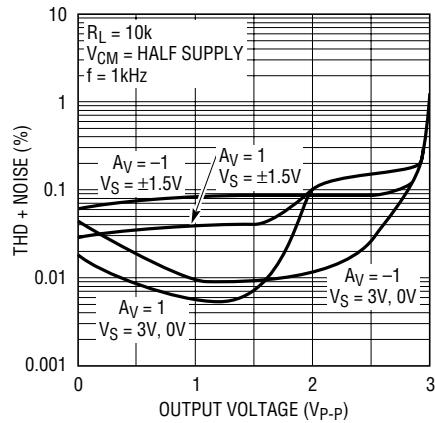
1490/91 G22

Total Harmonic Distortion + Noise vs Load Resistance



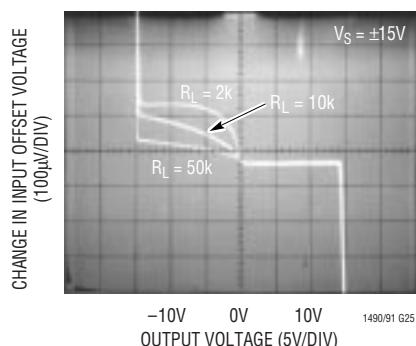
1490/91 G23

Total Harmonic Distortion + Noise vs Output Voltage



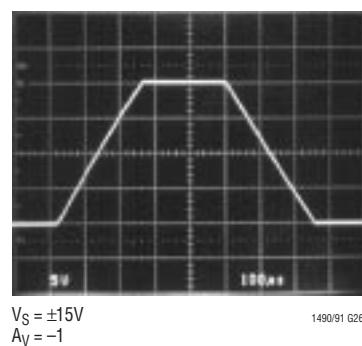
1490/91 G24

Open-Loop Gain



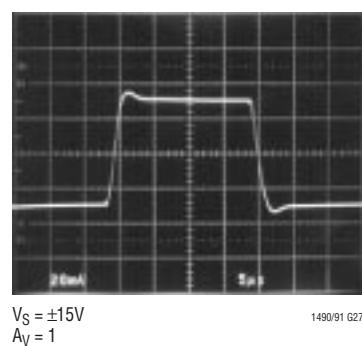
1490/91 G25

Large-Signal Response



1490/91 G26

Small-Signal Response



1490/91 G27

APPLICATIONS INFORMATION

Supply Voltage

The positive supply pin of the LT1490/LT1491 should be bypassed with a small capacitor (about $0.01\mu F$) within an inch of the pin. When driving heavy loads an additional $4.7\mu F$ electrolytic capacitor should be used. When using split supplies, the same is true for the negative supply pin.

The LT1490/LT1491 are protected against reverse battery voltages up to 18V. In the event a reverse battery condition occurs, the supply current is less than 1nA.

The LT1490/LT1491 can be shut down by removing V^+ . In this condition the input bias current is less than 0.1nA, even if the inputs are 44V above the negative supply.

When operating the LT1490/LT1491 on total supplies of 30V or more, the supply must not be brought up faster than $1\mu s$. This is especially true if low ESR bypass capacitors are used. A series RLC circuit is formed from the supply lead inductance and the bypass capacitor. 5Ω of resistance in the supply or the bypass capacitor will dampen the tuned circuit enough to limit the rise time.

Inputs

The LT1490/LT1491 have two input stages, NPN and PNP (see the Simplified Schematic), resulting in three distinct operating regions as shown in the Input Bias Current vs Common Mode typical performance curve.

For input voltages about 0.8V or more below V^+ , the PNP input stage is active and the input bias current is typically $-4nA$. When the input voltage is about 0.5V or less from V^+ , the NPN input stage is operating and the input bias current is typically $18nA$. Increases in temperature will cause the voltage at which operation switches from the PNP stage to the NPN stage to move towards V^+ . The input offset voltage of the NPN stage is untrimmed and is typically $600\mu V$.

A Schottky diode in the collector of each NPN transistor of the NPN input stage allows the LT1490/LT1491 to operate with either or both of its inputs above V^+ . At about 0.3V above V^+ the NPN input transistor is fully saturated and the input bias current is typically $4\mu A$ at room temperature. The input offset voltage is typically $700\mu V$ when operating above V^+ . The LT1490/LT1491 will operate with its inputs 44V above V^- regardless of V^+ .

The inputs are protected against excursions as much as 22V below V^- by an internal 1k resistor in series with each input and a diode from the input to the negative supply. There is no output phase reversal for inputs up to 22V below V^- . There are no clamping diodes between the inputs and the maximum differential input voltage is 44V.

Output

The output voltage swing of the LT1490/LT1491 is affected by input overdrive as shown in the typical performance curves. When monitoring voltages within 100mV of either rail, gain should be taken to keep the output from clipping.

The output of the LT1490/LT1491 can be pulled up to 18V beyond V^+ with less than 1nA of leakage current, provided that V^+ is less than 0.5V.

The normally reverse-biased substrate diode from the output to V^- will cause unlimited currents to flow when the output is forced below V^- . If the current is transient and limited to 100mA, no damage will occur.

The LT1490/LT1491 is internally compensated to drive at least 200pF of capacitance under any output loading conditions. A $0.22\mu F$ capacitor in series with a 150Ω resistor between the output and ground will compensate these amplifiers for larger capacitive loads, up to 10,000pF, at all output currents.

Distortion

There are two main contributors of distortion in op amps: output crossover distortion as the output transitions from sourcing to sinking current and distortion caused by nonlinear common mode rejection. Of course, if the op amp is operating inverting there is no common mode induced distortion. When the LT1490 switches between input stages there is significant nonlinearity in the CMRR. Lower load resistance increases the output crossover distortion, but has no effect on the input stage transition distortion. For lowest distortion the LT1490/LT1491 should be operated single supply, with the output always sourcing current and with the input voltage swing between ground and $(V^+ - 0.8V)$. See the Typical Performance Characteristics curves.

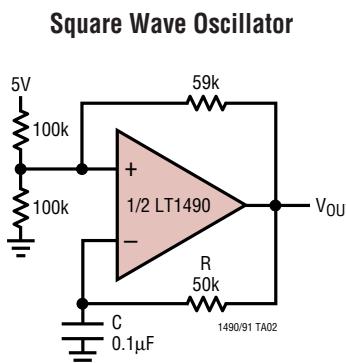
APPLICATIONS INFORMATION

Gain

The open-loop gain is almost independent of load when the output is sourcing current. This optimizes perfor-

mance in single supply applications where the load is returned to ground. The typical performance photo of Open-Loop Gain for various loads shows the details.

TYPICAL APPLICATIONS



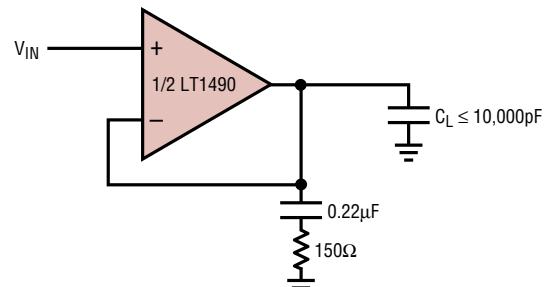
$$f = \frac{1}{2RC}$$

V_{OUT} = 5V_{P-P} WITH 5V SUPPLY

I_S = 200μA

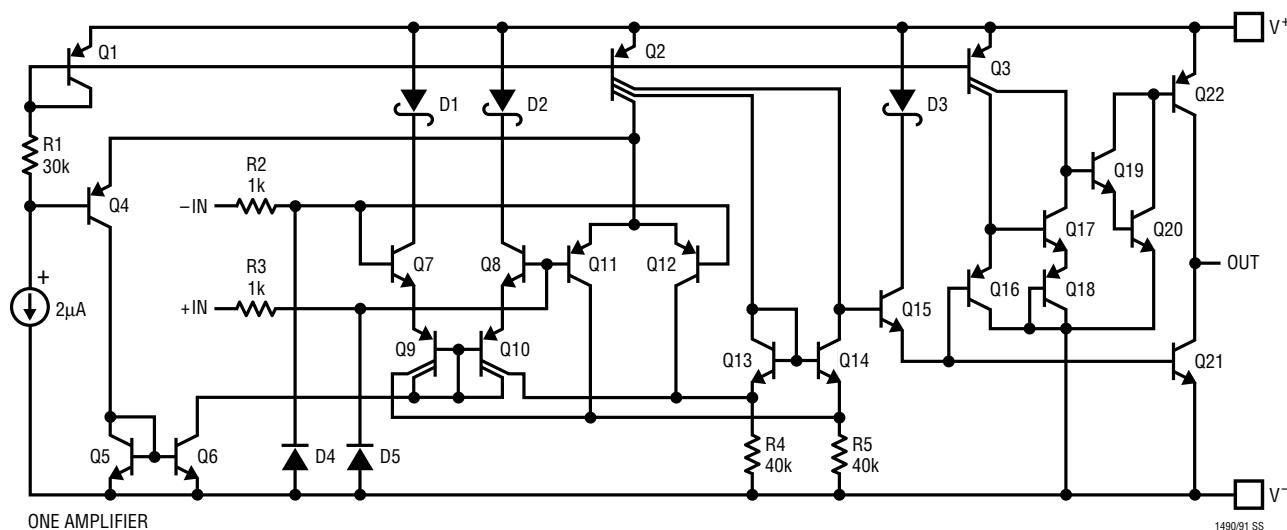
AT V_S = 5V, R = 50k, C = 1nF
OUTPUT IS 5kHz SLEW LIMITED TRIANGLE WAVE

Optional Output Compensation for Capacitive Loads Greater Than 200pF



1490/91 TA04

SIMPLIFIED SCHEMATIC

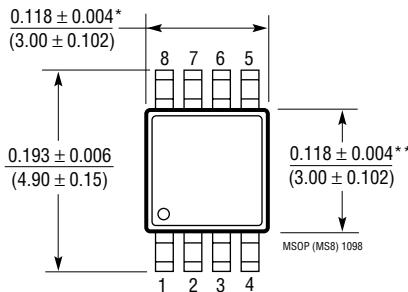
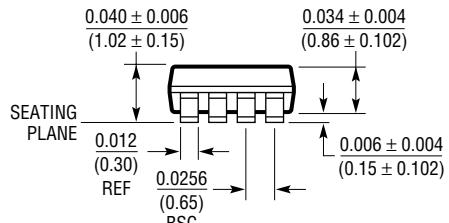
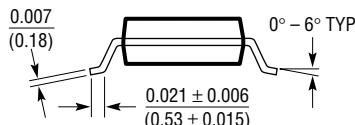


1490/91 SS

PACKAGE DESCRIPTION

Dimensions in inches (millimeters) unless otherwise noted.

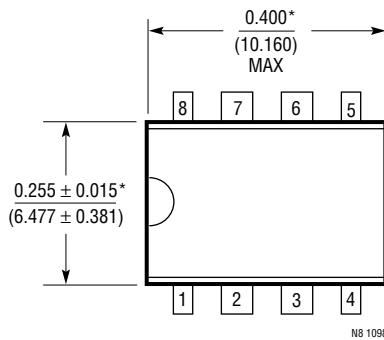
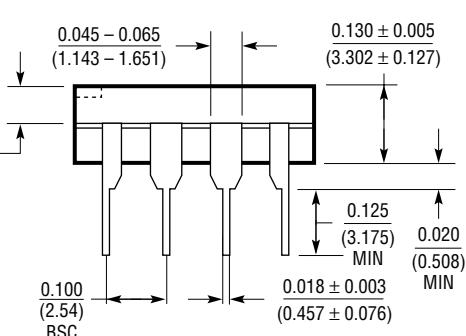
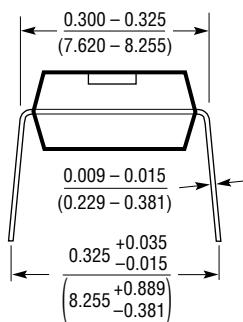
MS Package
8-Lead Plastic MSOP
(LTC DWG # 05-08-1660)



* DIMENSION DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

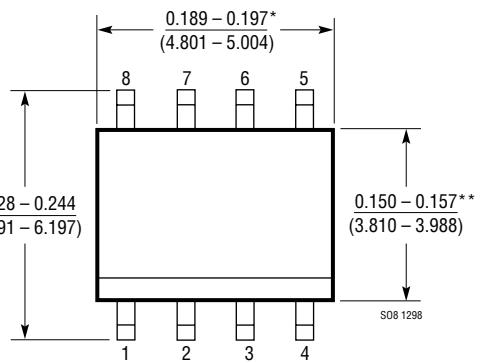
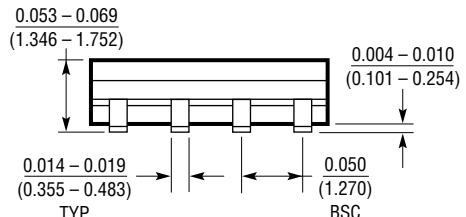
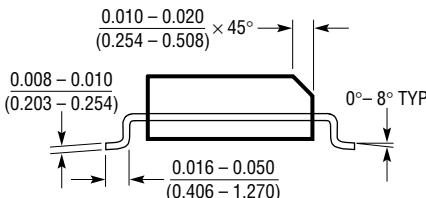
** DIMENSION DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
INTERLEAD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

N8 Package
8-Lead PDIP (Narrow 0.300)
(LTC DWG # 05-08-1510)



*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

S8 Package
8-Lead Plastic Small Outline (Narrow 0.150)
(LTC DWG # 05-08-1610)



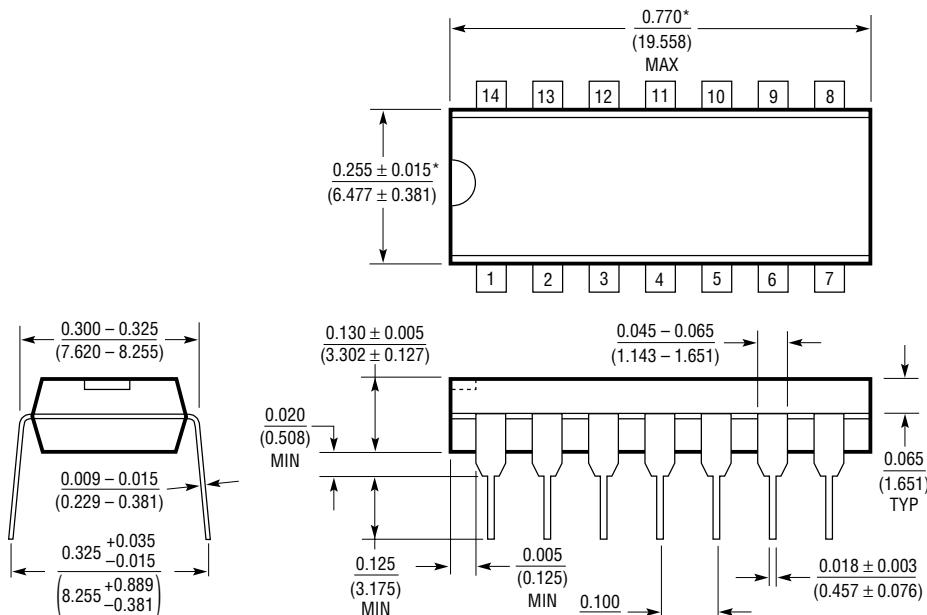
*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH
SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD
FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

PACKAGE DESCRIPTION

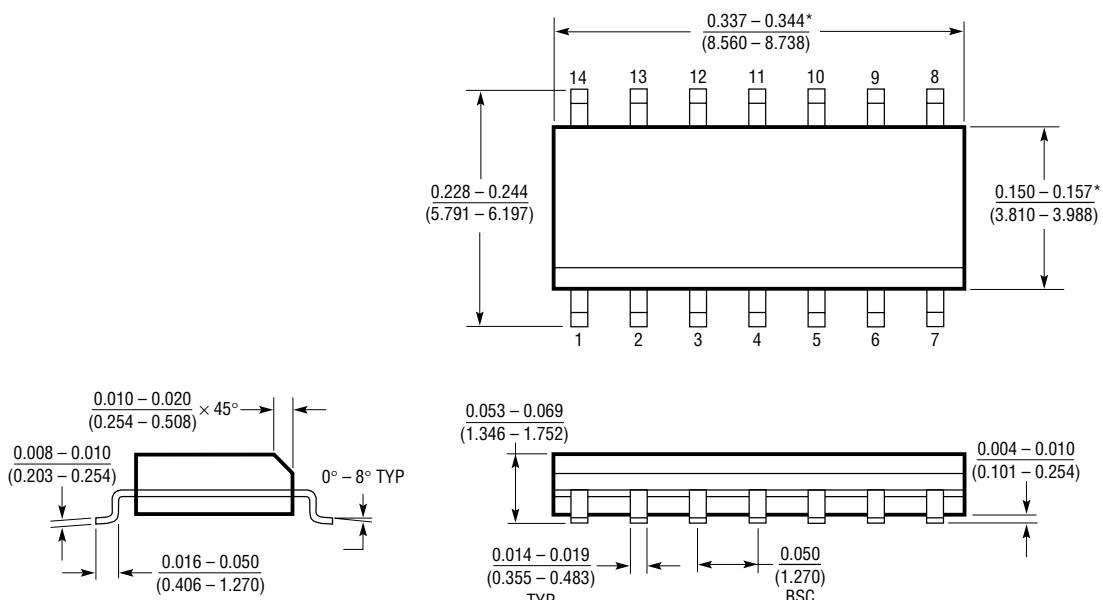
Dimensions in inches (millimeters) unless otherwise noted.

N Package
14-Lead PDIP (Narrow 0.300)
(LTC DWG # 05-08-1510)



N14 1098

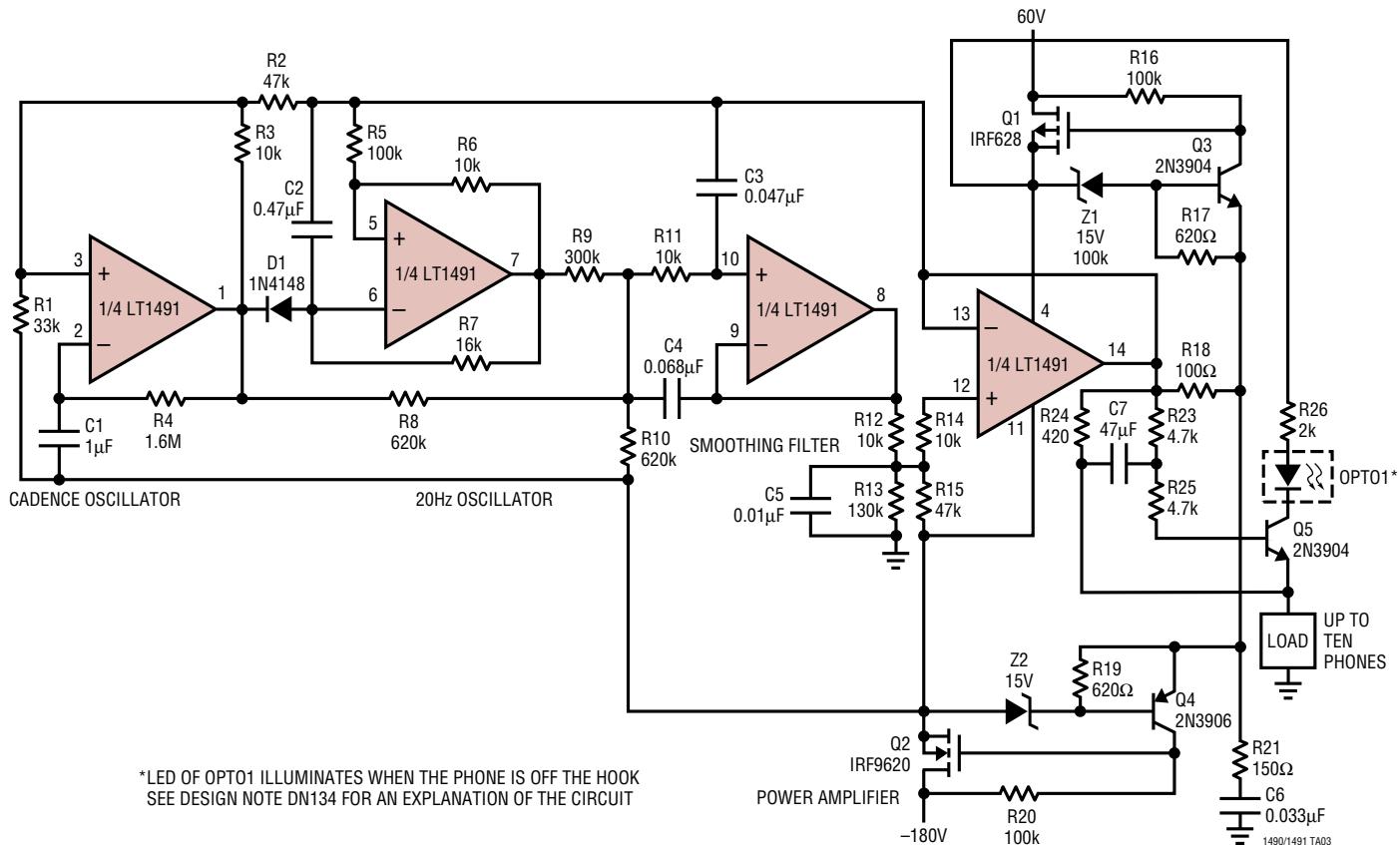
S Package
14-Lead Plastic Small Outline (Narrow 0.150)
(LTC DWG # 05-08-1610)



S14 1298

TYPICAL APPLICATION

Ring-Tone Generator



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1078/LT1079	Dual/Quad 55 μ A Max, Single Supply, Precision Op Amps	Input/Output Common Mode Includes Ground, 70 μ V V _{OS(MAX)} and 2.5 μ V/ $^{\circ}$ C Drift (Max), 200kHz GBW, 0.07V/ μ s Slew Rate
LTC1152	Rail-to-Rail Input, Rail-to-Rail Output, Zero-Drift Amplifier	High DC Accuracy, 10 μ V V _{OS(MAX)} , 100nV/ $^{\circ}$ C, 1MHz GBW, 1V/ μ s Slew Rate, Supply Current 2.2mA (Max), Single Supply, Can Be Configured for C-Load TM Operation
LT1178/LT1179	Dual/Quad 17 μ A Max, Single Supply, Precision Op Amps	Input/Output Common Mode Includes Ground, 70 μ V V _{OS(MAX)} and 4 μ V/ $^{\circ}$ C Drift (Max), 85kHz GBW, 0.04V/ μ s Slew Rate
LT1366/LT1367	Dual/Quad Precision, Rail-to-Rail Input and Output Op Amps	475 μ V V _{OS(MAX)} , 500V/mV A _{VOL(MIN)} , 400kHz GBW
LT1490A/LT1491A	Dual/Quad Micropower Rail-to-Rail Precision Op Amps	500 μ V V _{OS(MAX)} , Improved Version of the LT1490/LT1491

C-Load is a trademark of Linear Technology Corporation.