

Crimzon® Infrared Microcontrollers

ZLF645 Series Flash MCUs with Learning Amplification

Product Specification

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Revision History

Each instance in the revision history table reflects a change to this document from its previous revision. For more details, refer to the corresponding pages or appropriate links given in the table below.

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Architectural Overview

Maxim's ZLF645 Series of Flash MCU's are members of the Crimzon[®] family of infrared microcontrollers. This series provides a directly-compatible code upgrade path to other Crimzon MCUs, offers a robust learning function, and features up to 64 KB Flash memory and 1K general-purpose Random Access Memory (RAM). Two timers allow the generation of complex signals while performing other counting operations.

A Universal Asynchronous Receiver/Transmitter (UART) allows the ZLF645 MCU to function as a slave/master database chip. When the UART is not in use, the Baud Rate Generator (BRG) can be used as a third timer. Enhanced Stop Mode Recovery features allow the ZLF645 MCU to recover from STOP mode on any change of logic and on any combination of the 12 SMR inputs. The SMR source can also be used as an interrupt source.

Many high-end remote control units offer a learning function. A learning function allows a replacement remote unit to learn infrared signals from the original remote unit and regenerate the signal. However, the amplifying circuits of many learning remotes are expensive and are not tuned well. The ZLF645 MCU is the first chip to offer a built-in tuned amplification circuit in a wide range of positions and battery voltages. The only external component required is a photodiode.

The ZLF645 MCU greatly reduces the system cost and improves learning function reliability. With all new features, the ZLF645 MCU is excellent for infrared remote control and other MCU applications.

Features

Table 1 lists the memory, I/O, and power features of the ZLF645 Flash MCU. Additional features are listed below the table.

Table 1. ZLF645 Flash MCU Features

Interrupt Sources

The ZLF645 MCU supports 23 interrupt sources with 6 interrupt vectors, as given below:

- Three external interrupts.
- **•** Two from T8, T16 time-out and capture.
- **•** Three from UART Tx, UART Rx, and UART BRG.
- **•** One from LVD.
- **•** Fourteen from SMR source P20-P27, P30-P33, P00, and P07:
	- **–** Any change in logic from P20-P27, P30-P33 can generate an interrupt or SMR

Additional Features

The additional features of ZLF645 MCU include:

- **•** IR learning amplifier.
- Low power consumption—11 mW (typical).
- Three standby modes:
	- **–** STOP—1.7 A (typical)
	- **–** HALT—0.6 mA (typical)
	- **–** Low-voltage reset
- **•** Intelligent counter/timer architecture to automate generation or reception and demodulation of complex waveform, and pulsed signals:
	- **–** One programmable 8-bit counter/timer with two capture registers and two load registers
	- **–** One programmable 16-bit counter/timer with one 16-bit capture register pair and one 16-bit load register pair
	- **–** Programmable input glitch filter for pulse reception
	- **–** The UART baud rate generator can be used as another 8-bit timer, when the UART is not in use
- **•** Six priority interrupts:
	- **–** Three external/UART interrupts
	- **–** Two assigned to counter/timers
	- **–** One low-voltage detection interrupt

- **•** 8-bit UART:
	- $-$ R_x and T_x interrupts
	- **–** 4800, 9600, 19200, and 38400 baud rates
	- **–** Parity Odd/Even/None
	- **–** Stop bits 1/2
- **•** ICP (In-circuit Flash Programming) interface multiplexed with one of the GPIO's.
- Intelligent Power-On Reset (POR) to provide reduced POR time on detection of stable clock from external crystal oscillator or resonator.
- **•** Low-voltage and high-voltage detection flags.
- **•** Programmable Watchdog Timer (WDT)/POR circuits.
- **•** Two on-board analog comparators with independent reference voltages and programmable interrupt polarity.
- **•** User-selectable options through option bit Flash coding (ON/OFF):
	- **–** Port 0 pins 0–3 pull-up transistors
	- **–** Port 0 pins 4–7 pull-up transistors
	- **–** Port 1 pins 0–3 pull-up transistors
	- **–** Port 1 pins 4–7 pull-up transistors
	- **–** Port 2 pins 0–7 pull-up transistors
	- **–** Port 3 pins 0–3 pull-up transistors
	- **–** Port 4 pins 0–7 pull-up transistors
	- **–** WDT enabled at Power-On Reset
	- **–** Flash lowest half main memory protect
	- **–** Flash entire main memory protect
	- **–** 16-bit addressability for stack pointer
	- **–** No division, divide by 2, divide by 16, or divide by 32 of external clock to system clock

All signals with an overline, " ", are active Low. For example, B/W, in which WORD is active Low, and B/W, in which BYTE is active Low. **Note:**

Power connections use the conventional descriptions listed in Table 2.

Table 2. Power Connections

Connection	Device
Power	Vnn
Ground	V_{SS}

Functional Block Diagram

Figure 1 displays the functional blocks of the ZLF645 Flash MCU.

Pin Description

Figure 2 displays the pin configuration for ZLF645 MCU 20-pin QFN packages.

Figure 2. 20-Pin QFN Pin Configuration

Table 3 lists the function and signal directions of each pin within the 20-pin QFN package sequentially by pin number.

Table 3. 20-Pin QFN Sequential Pin Identification

through the pull-up to Ground.

Table 4 lists the function and signal direction of each pin within the 20-pin QFN package by function.

Table 4. 20-Pin QFN Functional Pin Identification

Note: When the Port 0 low-nibble pull-up option is enabled and the P30 input is Low, current flows through the pull-up to Ground.

Figure 3 displays the pin configuration for ZLF645 MCU 20-pin PDIP, SOIC, and SSOP packages.

Figure 3. 20-Pin PDIP/SOIC/SSOP Pin Configuration

Table 5 lists the function and signal directions of each pin within the 20-pin PDIP, SOIC, and SSOP packages sequentially by pin number.

Table 5. 20-Pin PDIP/SOIC/SSOP Sequential Pin Identification

Note: When the Port 0 low-nibble pull-up option is enabled and the P30 input is Low, current flows through the pull-up to Ground.

Table 6 lists the function and signal direction of each pin within the 20-pin PDIP, SOIC, and SSOP packages by function.

Table 6. 20-Pin PDIP/SOIC/SSOP Functional Pin Identification

Note: When the Port 0 low-nibble pull-up option is enabled and the P30 input is Low, current flows through the pull-up to Ground.

Figure 4 displays the pin configuration of the ZLF645 MCU within the 28-pin PDIP, SOIC, and SSOP packages.

	$P25\Box$ $P26\Box$ $P27\Box$ $P04\square$ $P05\Box$ $P06\Box$ $PO7$ \square Voo⊑ XTAL2 □ XTAL1 □ $P31\Box$ $P32\Box$ $P33\Box$ $P34\Box$	1 2 3 4 5 6 7 8 9 10 11 12 13 14	28-Pin PDIP SOIC SSOP	28 27 26 25 24 23 22 21 20 19 18 17 16 15	\square P24 \square P23 \square P22 ⊐ P21 ⊐ P20 \square P03 v_{ss} ⊒ P02 ⊒ P01 ⊐ P00 \square P30 ⊐ P36 ⊐ P37 ⊐ P35		
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Figure 4. 28-Pin PDIP/SOIC/SSOP Pin Configuration

Table 7 lists the function and signal directions of each pin within the 28-pin PDIP, SOIC, and SSOP packages sequentially by pin number.

Table 8 lists the functions and signal directions of each pin within the 28-pin PDIP, SOIC, and SSOP packages by function.

Table 8. 28-Pin PDIP/SOIC/SSOP Functional Pin Identification

Figure 5 displays the pin configuration of the ZLF645 MCU within the 48-pin SSOP package.

Figure 5. 48-Pin SSOP Pin Configuration

Table 9 lists the functions and signal directions of each pin within the 48-pin SSOP package sequentially by pin number.

Table 9. 48-Pin SSOP Sequential Pin Identification (Continued)

Table 9. 48-Pin SSOP Sequential Pin Identification (Continued)

Table 10 lists the functions and signal directions of each pin within the 48-pin SSOP package by function.

Table 10. 48-Pin SSOP Functional Pin Identification

Table 10. 48-Pin SSOP Functional Pin Identification (Continued)

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I/O Port Pin Functions

The ZLF645 MCU features up to five 8-bit ports which are described below:

- 1. Port 0 is nibble-programmable as either input or output.
- 2. Port 1 is byte-programmable as either input or output.
- 3. Port 2 is bit-programmable as either input or output.
- 4. Port 3 features four inputs on the lower nibble and four outputs on the upper nibble.
- 5. Port 4 is bit-programmable as either input or output.

Port 0, Port 1, Port 2, and Port 4 internal pull-ups are disabled on any pin or group of pins when programmed into output mode. **Note:**

The CMOS input buffer for each Port 0, Port 1, Port 2, or Port 4 pin are always **Caution:***connected to the pin, even when the pin is configured as an output. If the pin is configured as an open-drain output and no external signal is applied, a High output state can cause the CMOS input buffer to float. This may lead to excessive leakage current of more than 100 A. To prevent this leakage, connect the pin to an external signal with a defined logic level or ensure that its output state is Low, especially during STOP mode.*

> *Port 0, Port 1, Port 2, and Port 4 have both input and output capability. The input logic is always present no matter whether the port is configured as input or output. When executing a READ instruction, the MCU reads the actual value at the input logic but not from the output buffer. In addition, the instructions of OR, AND, and XOR have the Read-Modify-Write sequence. The MCU first reads the port, then modifies the value, and loads back to the port.*

> *Precaution must be taken, if the port is configured as an open-drain output or if the port is driving any circuit that makes the voltage different from the appropriate output logic. If it is configured as open-drain output with output logic as ONE, it is a floating port and reads back as ZERO. The following instruction sets P00–P07 all Low:*

AND P0,#%F0

RESET (Input, Active Low)

Reset initializes the MCU and is accomplished either through Power-On Reset (POR), Watchdog Timer (WDT), Stop Mode Recovery, Low-Voltage detection, or through the external reset pin in the case of 48-pin packaged products.

During POR and WDT Reset, the internally generated reset drives the reset pin Low for the POR time. Any device driving the external reset line must be open-drain to avoid damage from a possible conflict during reset conditions. A pull-up is provided internally for the reset pin, if available. When the ZLF645 MCU asserts (Low) the RESET pin, the internal pull-up is disabled. The ZLF645 MCU does not assert the RESET pin when the VDD voltage is below the VBO trip point level (for more details, see Reset and Power Management on page 137).

▶ **Note:***The external reset does not initiate an exit from STOP mode.*

> Table 11 lists the registers used to control I/O ports. Some port pin functions can also be affected by control registers for other peripheral functions.

Table 11. I/O Port Control Registers

Port 0

Port 0 is an 8-bit bidirectional CMOS-compatible port. Its eight I/O lines are configured under software control to create a nibble I/O port. The output drivers are push/pull or open-drain, controlled by bit 2 of the Port Configuration Register.

If one or both nibbles are required for I/O operation, they must be configured by writing to the Port 0/1 Mode Register. After a hardware reset or a Stop Mode Recovery, Port 0 is configured as an input port.

Port 0, bit 7 is used as the transmit output of the UART when UART Tx is enabled. The I/O function of Port 0, bit 7 is overridden by the UART serial output (TxD) when UART Tx is enabled (UCTL[7] = 1). The pin must be configured as an output for TxD data to reach the pin $(PO1M[6] = 0)$.

An optional pull-up transistor is available as an user-selectable flash programming option on all Port 0 bits with nibble select. Figure 6 displays the Port 0 configuration.

Figure 6. Port 0 Configuration

Port 1

Port 1 is an 8-bit bidirectional CMOS-compatible I/O port. It can be configured under software control as inputs or outputs. A flash programming option bit is available to connect eight pull-up transistors on this port. Bits programmed as output are globally programmed as either push/pull or open-drain. The power-on reset function resets with the eight bits of Port 1 [P17:10] configured as inputs. Figure 7 displays the Port 1 configuration.

Port 2

Port 2 is an 8-bit bidirectional CMOS-compatible I/O port. Its eight I/O lines can be independently configured under software control as inputs or outputs. Port 2 is always available for I/O operation. A flash programming option bit is available to connect eight pull-up transistors on this port. Bits programmed as outputs are globally programmed as either push/pull or open-drain. The Power-On Reset function resets with the eight bits of Port 2 [P27:20] configured as inputs.

Port 2 also has an 8-bit input OR and AND gate and edge detection circuitry, which can be used to recover from the STOP mode. P20 can be programmed to access the edge-detection circuitry in DEMODULATION mode. Figure 8 displays the Port 2 configuration.

Figure 8. Port 2 Configuration

Port 3

Port 3 is an 8-bit CMOS-compatible I/O port. Port 3 consists of four fixed inputs (P33:P30), three fixed outputs (P37:P36:P35), and one multi-functioned pin (P34) that can function as an output only or as a bidirectional open-drain I/O depending on whether the ZLF645 MCU is in ICP mode.

P30, P31, P32, and P33 are standard CMOS inputs with option enabled pull-up transistors and can be configured under software as interrupts, as received data input to the UART block, as input to comparator circuits, or as input to the IR learning amplifier.

P37, P36, and P35 are push/pull outputs and can be configured as outputs from counter/timers and/or comparator circuits.

During the ZLF645's POR time, P34 is configured as an input pin with pull-up enabled. If after completing it's POR period, the ZLF645 has not detected this pin LOW and been put into ICP mode, this pin will revert back to being a push/pull output only. For more details on the function of pin P34, see ICP Interface on page 53.

Figure 9 displays the Port 3 configuration.

P31 can be used as an interrupt, analog comparator input, infrared learning amplifier input, normal digital input pin, and as a Stop Mode Recovery source. When bit 2 of the Port 3 Mode register (P3M) is set, P31 is used as the infrared learning amplifier, IR1. The reference source for IR1 is GND. The infrared learning amplifier is disabled during STOP mode. When bit 1 of P3M is set, the part is in ANALOG mode and the analog comparator,

COMP1 is used. The reference voltage for COMP1 is P30 (P_{REF1}). When in ANALOG mode, P30 cannot be read as a digital input when the CPU reads bit 0 of the Port 3 register; such reads always return a value of 1.

Also, when in ANALOG mode, P31 cannot be used as a Stop Mode Recovery source, as in STOP mode the comparator is disabled and its output will not toggle. The programming of bit 2 of the P3M register takes precedence over the programming of Bit 1 in determining the function of P31. If both bits are set, P31 functions as an IR learning amplifier instead of an analog comparator. As displayed in Figure 9 the output of the function selected for P31 can be used as a source for IRQ2 interrupt assertion. The IRQ2 interrupt can be configured based upon detecting a rising, falling, or edge-triggered input change using Bits 6 and 7 of the IRQ register. The P31 output stage signal also goes to the Counter/Timer edge detection circuitry in the same way that P20 does.

P32 can be used as an interrupt, analog comparator, UART receiver, normal digital input and as Stop Mode Recovery source. When bit 6 of UCTL register is set, P32 functions as a receive input for the UART. When bit 1 of the P3M register is set, thereby placing Port 3 into ANALOG mode, P32 functions as an analog comparator, COMP2. The reference voltage for COMP2 is P33 (P_{REF2}). P32 can be used as a rising, falling or edge-triggered interrupt, IRQ0, using IRQ register bits 6 and 7. If UART receiver interrupts are not enabled, the UART receive interrupt is used as the source of interrupts for IRQ0 instead of P32. When in ANALOG mode P32 cannot be used as SMR source because the comparators are turned OFF in STOP mode.

When in ANALOG mode, P33 cannot be read by the CPU as a digital input through bit 3 of the Port 3 register. In this case, a read of bit 3 of the Port 3 register indicates whether Stop Mode Recovery condition exists. Reading a value of 0 indicates an SMR condition; if the ZLF645 MCU is in STOP mode, it will exit STOP mode. Reading a value of 1 indicates that no condition exists to exit the ZLF645 MCU from STOP mode.

Additionally, when in ANALOG mode, P33 cannot be used as an interrupt source. Instead, the existence of a SMR condition can generate an interrupt, if enabled. P33 can be used as a falling-edge interrupt, IRQ1, when not in ANALOG mode. IRQ1 is also used as the UART T_X interrupt and the UART BRG interrupt. Only one source is active at a time. If bit 7 and bit 5 of UCTL are set to 1, IRQ1 will transmit an interrupt when the Transmit Shift register is empty. If bits 0 and 5 of UCTL are set to 1 and bit 6 of UCTL is cleared to 0, the BRG interrupts will activate IRQ1.

Comparators and the IR amplifier are powered down by entering STOP mode. **Note:**

> *For P30:P33 to be used as a Stop Mode Recovery source during STOP mode, these inputs must be placed into DIGITAL mode. When in ANALOG mode, do not configure any Port 3 input as a SMR source. The configuration of these inputs must be re-initialized after Stop Mode Recovery or POR.*

Table 12.Summary of Port 3 Pin Functions

Port 3 also provides output for each of the counter/timers and AND/OR Logic (see Figure 10). Control is performed by programming CTR1 bit 5 and bit 4, CTR0 bit 0, and CTR2 bit 0.

Figure 10. Port 3 Counter/Timer Output Configuration

Comparator Inputs

In ANALOG mode, P31 and P32 have a comparator front end. The comparator reference is supplied by P33 and P_{REF1} . In ANALOG mode, the P33 internal data latch and its corresponding IRQ1 are diverted to the Stop Mode Recovery sources (excluding P31, P32, and P33) as displayed in Figure 9 on page 24. In DIGITAL mode, P33 is used as bit 3 of the Port 3 input register, which then generates IRQ1.

Comparators are powered down by entering STOP mode. For P30:P33 to be used as an SMR source, these inputs must be placed into DIGITAL mode. **Note:**

Comparator Outputs

The comparators can be programmed to output on P34 and P37 by setting bit 0 of the PCON register.

Port 4

Port 4 is an 8-bit bidirectional CMOS-compatible I/O port. Its eight I/O lines can be independently configured under software control as inputs or outputs. Port 4 is always available for I/O operation. A flash programming option bit is available to connect eight pull-up transistors on this port. Bits programmed as outputs are globally programmed as either push/pull or open-drain. The POR function resets with the eight bits of Port 4 [P47:40] configured as inputs. Figure 11 on page 29 displays the Port 4 configuration.

Port Configuration Register

The Port Configuration register (see Table 13) configures the Port 0 output mode and the comparator output on Port 3. The PCON register is located in expanded register Bank F, address 00h.

Table 13. Port Configuration Register (PCON)

PCON register is not reset after a Stop Mode Recovery. Also, for package types other than the 48-pin package, writes to bit 3 and bit 1 have no effect. **Note:**

 \blacktriangleright

Port 0/1 Mode Register

The Port 0/1 Mode register (see Table 14) determines the I/O direction of Port 0 and Port 1. The Port 0 direction is nibble-programmable. Bit 6 controls the upper nibble of Port 0, bits [7:4]. Bit 0 controls the lower nibble of Port 0, bits [3:0]. The Port 1 direction is byte programmable.

Table 14. Port 0/1 Mode Register (P01M)

Bit		6	5	4		2			
Field	Reserved	P07:P04 Mode	Reserved		Port 1 Mode	Reserved		P03:P00 Mode	
Reset			х	X		x			
R/W		W			W			W	
Address	Bank Independent: F8h; Linear: 0F8h								

* For package types other than the 48-pin package, writes to bit 3 have no effect.

Only P00, P01, and P07 are available for ZLF645 Flash MCU 20-pin configuration. **Note:**

Port 0 Register

The Port 0 register (see Table 15) allows read and write access to the Port 0 pins.

Table 15. Port 0 Register (P0)

▶

Only P00, P01, and P07 are available for ZLF645 Flash MCU 20-pin configuration. **Note:**

Port 1 Register

The Port 1 register (see Table 16) allows read and write access to the Port 1 pins.

Table 16. Port 1 Register (P1)

Port 2 Mode Register

The Port 2 Mode register (see Table 17) determines the I/O direction of each bit on Port 2. Bit 0 of the Port 3 Mode register determines whether the output drive is push/pull or open-drain.

Table 17. Port 2 Mode Register (P2M)

Bit		6			3	2			
Field	P27 I/O Definition	P ₂₆ I/O Definition	P ₂₅ I/O Definition	P ₂₄ I/O Definition	P ₂₃ I/O Definition	P ₂₂ I/O Definition	P21 I/O Definition	P20 I/O Definition	
Reset									
R/W	w	w	W	W	W	w	W	W	
Address	Bank Independent: F6h; Linear: 0F6h								

Note:

Port 2 Mode register is not reset after a Stop Mode Recovery.

Port 2 Register

The Port 2 register (see Table 18) allows read and write access to the Port 2 pins.

Table 18. Port 2 Register (P2)

Port 3 Mode Register

The Port 3 Mode register (see Table 19) is used to configure the functionality of Port 3 inputs and the output mode of Port 2. When bit 2 is set, the IR Learning Amplifier is used instead of the COMP1 comparator, regardless of the value of bit 1.

Table 19. Port 3 Mode Register (P3M)

Port 3 Mode register is not reset after a Stop Mode Recovery. **Note:**

Port 3 Register

The Port 3 register (see Table 20) allows read access to port pins P33 through P30 and write access to the port pins P37 through P34.

Table 20. Port 3 Register (P3)

Port 3 register is not reset after a Stop Mode Recovery. **Note:**

Port 4 Mode Register

The Port 4 Mode register (see Table 21) determines the I/O direction of each bit on Port 4. Bit 3 of the Port Configuration register (PCON) determines whether the output drive is push/pull or open-drain.

Table 21. Port 4 Mode Register (P4M)

Bit		6		4	3	2			
Field	P47 I/O Definition	P46 I/O Definition	P45 I/O Definition	P44 I/O Definition	P43 I/O Definition	P42 I/O Definition	P41 I/O Definition	P40 I/O Definition	
Reset									
R/W	W	W	W	W	W	W	W	W	
Address	Bank F: 09h; Linear: F09h								

Port 4 Mode register is not reset after a Stop Mode Recovery. **Note:**

Port 4 Register

The Port 4 register (see Table 22) allows read and write access to the Port 4 pins.

Table 22. Port 4 Register (P4)

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Memory and Registers

The Z8® LXMC CPU used in the ZLF645 Series of Flash MCUs incorporates special features to extend the available memory space while maintaining the benefits of a $Z8^{\circledast}$ CPU core in battery-operated applications.

Flash Program/Constant Memory

The ZLF645 Series of Flash MCUs can address up to 64 KB of Flash memory for object code (program instructions and immediate data) and constant data (ROM tables and data constants). The first 12 bytes of the memory are reserved for the six available 16-bit interrupt request (IRQ) vectors. On reset, program execution begins at address 000Ch in the memory. Execution rolls over to the beginning of the memory if the program counter address exceeds the Flash memory size.

The entire Flash memory is available for either program code or constant data. Outside of normal instruction fetches, the CPU can access the Flash memory by using LDC and LDCI instructions. The LDC and LDCI instructions use 16-bit addresses to access the memory. Figure 12 displays the Program/Constant memory map for the device.

0000h = 16-bit Address **(Not to Scale)**

Figure 12. Program/Constant Memory Map

Register File

The ZLF645 Series of Flash MCUs features up to 1024 bytes of register file space, organized in 256-byte banks. Bank 0 contains 235 or 237 bytes of RAM addressed as general purpose registers, 5 or 3 port addresses, and 16 control register addresses. For 20- or 28-pin packages, Port 1 and Port 4 registers of Bank 0 are not implemented and there locations are available as general-purpose registers. Bank 1, Bank 2, and Bank 3; each contain 256 general-purpose register bytes. Bank D and Bank F; each contain 16 addresses for control registers. All other banks are reserved and must not be selected.

The current bank is selected for 8-bit direct or indirect addressing by writing Register Pointer bits RP[3:0]. In the current bank, a 16-byte working register group (addressed as $R0-R15$) is selected by writing RP[7:4]. A working register operand requires only 4 bits of Program Memory. There are 16 working register groups per bank (see Figure 13 and Figure 14).

The 8-bit addresses in the range F0h–FFh (and the equivalent 4-bit addresses) are bank-independent, meaning they always access the control registers in Bank 0, regardless of the RP[3:0] value. Addresses in the range 00h–03h always access the Bank 0 Port registers unless Bank D or Bank F is selected (Port 01h is not implemented in this device). When Bank D or Bank F is selected, addresses 10h–EFh access the Bank 0 generalpurpose registers.

The LDX and LDXI instructions or indirect addressing is used to access the Bank 1–3 registers not accessible by 8-bit or working register addresses (12-bit addresses—100h– 103h, 1F0h–1FFh, 200h–203h, 2F0h–2FFh, 300h–303h, and 3F0h–3FFh). See Linear Memory Addressing on page 45.

Stack

The Stack Pointer register provides either 16-bit or 8-bit of stack pointer addressability depending upon the programming of bit 3 of User Option Byte 1 (for more details, see Flash Option Bits on page 171).

16-bit Stack Addressability

When programmed for 16-bit stack addressability, the stack address is formed as a combination of the SPL and SPH registers located at addresses FFh and FEh. For 1K and 512 B RAM products, the most significant 6 or 5 bits, respectively of the SPH register are ignored. The stack address is mapped to a particular RAM memory location by the following formula:

 $Bank = \{2'b0, SPH[1:0]\}$

 $Group = SPI_[7:4]$

Register number = $SPL[3:0]$

With the ZLF645 MCU configured for 16-bit stack addressability, stack reads or writes to Bank 3, 2, 1, or 0 Group F Registers or to any of the Port registers actually accesses

shadow registers implemented within the RAM memory. This enables the entire 1K or 512 B, depending on the product, of the RAM memory to be used for the stack.

8-bit Stack Addressability

For 8-bit stack addressability, only the SPL register is used for stack addressing and stack operations that use the stack pointer always address Bank 0, independent of the RP[3:0] setting. For more details on the stack, refer to *Z8*® *LXMC CPU Core User Manual (UM0215)*.

When in 8-bit stack addressability mode, the Bank 0 register FEh can be used to store user data. See Stack Pointer Register on page 48.

** For 20 and 28 pin parts, the Port01 and Port04 locations become available for use as general purpose registers

Figure 13. Register File 8-Bit Banked Address Map

* RP=00: selects Register Bank 0, Working Register Group 0

Figure 14. Register Pointer—Detail

Register Pointer Example

The counter/timers are mapped into ERF Group D. Access is easily performed using the following code segment:

```
LD RP, #0Dh ; Select ERF D for access to Bank D
               ; (working register group 0)
LD R0, #xx ; load CTR0
LD 1, #xx \qquad ; load CTR1
LD R1, 2 \qquad ; CTR2 \rightarrow CTR1
LD RP, #7Dh ; Select Expanded Register Bank D and working
              ; register group 7 of Bank 0 for access.
LD 71h, 2 : CTR2 \rightarrow \text{register 71h}LD R1, 2 : \text{CTR2} \rightarrow \text{register} 71h
```
Linear Memory Addressing

In addition to using the RP register to designate a bank and working register group for 8-bit or 4-bit addressing, programs can use 12-bit linear addressing to load a register in any other bank to or from a register in the current bank. Linear addressing is implemented through the LDX and LDXI instructions only. Linear addressing treats the register file as if all the registers are logically ordered end-to-end, as opposed to being grouped into banks and working register groups, as displayed in Figure 15 on page 47. For linear addressing, register file addresses are numbered sequentially from Bank 0, register 00h to Bank 0, register FFh, then continuing with Bank 1, register 00h, and so on up to Bank F, register FFh.

Using the LDX and/or the LDXI instructions, either the target or destination register location can be addressed through a 12-bit linear address value stored in a general-purpose register pair.

Example

For example, the following code uses linear addressing for the source of a register transfer operation and uses a working register address for the target:

In the above code, the source register is referred through a linear address value contained within registers R6 and R7, whereas the destination is referenced via the SRP setting and a working register. For more information about instructions on the usage of LDX and LDXI instructions, refer to *Z8*® *LXMC CPU Core User Manual (UM0215)*.

The LDE and LDEI instructions that existed in the Z8 CPU are no longer valid; they have been replaced by the LDX and LDXI instructions. **Note:**

▶

** For 20 and 28 pin parts, the Port01 and Port04 locations become available for use as general purpose registers

Figure 15. Register File LDX, LDXI Linear 12-Bit Address Map

Register Pointer Register

The upper nibble of the Register Pointer register (see Table 23) selects which working register group is accessed. A working register group consists of 16 bytes. The lower nibble selects the expanded register file bank; for ZLF645 MCU, Banks 0, 1, 2, 3, F, and D are implemented. A 0h in the lower nibble allows the normal register file (Bank 0) to be addressed. Any other value from 01h to 0Fh exchanges the lower 16 registers to an expanded register bank.

Table 23. Register Pointer Register (RP)

Stack Pointer Register

Through a Flash programmable option bit, the Stack Pointer register of the ZLF645 MCU is either one or two bytes providing either 8-bit or 16-bit of stack addressing. When not enabled through the option bit for 16-bit stack addressability, the SPH register can be used as a User Data register (USER). The stack pointer resides in the RAM and when the ZLF645 MCU is programmed for 8-bit addressing, this stack pointer resides in Bank 0 of the RAM only. With 16-bit addressing, the entire RAM's address space is available for use as the stack.

The stack address is decremented prior to a PUSH operation and incremented after a POP operation. The stack address always points to the data stored at the 'top' of the stack (the lowest stack address). During a call instruction, the contents of the Program Counter are saved on the stack. Interrupts cause the contents of the Program Counter and Flags registers to be saved on the stack. An overflow or underflow can occur when the stack address is incremented or decremented during normal operations. You must prevent this occurrence or unpredictable operations may result (see Table 24 on page 49).

Table 24. Stack Pointer Register Low Byte (SPL)

Table 25.Stack Pointer Register High Byte (SPH) or User Data Register (USER)

Notes:

- 1. *For devices with 1K bytes of RAM and with 16-bit stack pointer mode enabled, the upper 6 bits of this register are unused for stack addressing. For devices with 512 bytes of RAM and with 16-bit stack pointer mode enabled, the upper 7 bits of this register are unused for stack addressing.*
	- *2. When ZLF645 MCU is not in 16-bit stack pointer mode, this register is available to store use user data and its functionality is identical to other Maxim® Crimzon products such as the ZLP12840 and ZLR64400 MCUs. When available for user data, this register must not be used as a counter for the DJNZ instruction.*

Register File Summary

Table 26 lists each linear (12-bit) register file address to the associated register, mnemonic, and reset value. The table also lists the register bank (or banks) and corresponding 8-bit address (if any) for each register and a page link to the detailed register table.

Throughout this document, an 'X' denotes an undefined digit. A ' \rightarrow ' (dash) in a table cell indicates that the corresponding attribute does not apply to the listed item. Reset value digits (highlighted in grey) are not reset by a Stop Mode Recovery. Register bit SMR[7] (shown in **boldface**) is set to 1 instead of reset by a Stop Mode recovery.

Table 26. Register File Address Summary

Address (Hex)

Table 26. Register File Address Summary (Continued)

Address (Hex)

Table 26. Register File Address Summary (Continued)

 $\overline{}$

1When ZLF645 is programmed for 16-bit stack addressability, the value in this register is used as the high byte of a 16-bit stack pointer.

ICP Interface

The ICP interface of the ZLF645 is a single pin RS-232 like interface for performing programming, reads, and memory erasures to the ZLF645's Flash memory. For enabling the ZLF645 into ICP mode and for performing ICP operations, the ZLF645's P34 pin which normally functions as an output only is used.

Enabling ICP Mode

As mentioned previously, the ZLF645's GPIO pin P34 is multi-functioned to be used for putting the ZLF645 into ICP mode and for ICP communications once it is in that mode. Entry into ICP mode takes place during the ZLF645's power on reset period. During the ZLF645's power on reset period, the P34 pin which normally is an output only pin is configured by the ZLF645 as an input with pull-up enabled. If during this time this pin is driven LOW and held LOW until the end of the power on reset period, the ZLF645 will be put into ICP mode. Once in ICP mode, the P34 pin operates as an open-drain output bidirectional pin with pull-up enabled. The power on reset period as can be seen from the electrical specs section of this document can have a duration range of between 2.5 ms and 10 ms. To ensure proper entry into ICP mode, the P34 pin should be driven LOW and held low a minimum of 10 ms after power up.

If during the ZLF645's power on reset period, the P34 pin is never driven LOW, pin p34 will be pulled HIGH through its pull-up device. In this case, if P34 remains HIGH until the end of the power on reset period, the ZLF645 will go into normal user mode and P34 will revert back to being an output pin only. To ensure proper entry into user mode when it is not intended to put the ZLF645 into ICP mode, it is important that in the customer application P34 only be connected to capacitive loads. This is due to the weak nature of its pull-up device, which can have a resistance ranging between 100 k Ω up to 600 k Ω depending on voltage, temperature, and process.

State of ZLF645 in ICP Mode

The operating characteristics of the device in ICP mode are:

- The CPU stops executing instructions.
- All on-chip peripherals are disabled.
- **•** The ZLF645 constantly refreshes the Watchdog Timer, if enabled.
- **•** The P34 pin is configured as a bidirectional pin with pull-up enabled and with the output stage configured as open-drain. The bidirectional control of the pins comes from the ICP Tx/Rx logic.

Enabling Flash Accesses Through the ICP

After the ZLF645 is in ICP mode, the FLASHCTL bit of the ICP Control register must be programmed to 1 before Flash accesses are enabled through the ICP interface.

ICP Interface Logic Architecture

The ICP logic within the ZLF645 MCU consists of four primary functional blocks: transmitter, receiver, auto-baud detector/generator, and Flash Controller interface. Figure 16 displays the architecture of the ICP.

Figure 16. In-Circuit Programmer Block Diagram

ICP Interface Operation

After the ZLF645 MCU is in ICP mode, pin P34 acts a bidirectional open-drain interface with internal pull-ups used for transmitting and receiving the data. Data transmission is half-duplex, in that transmit and receive cannot occur simultaneously. Serial data on P34 is sent using the standard asynchronous data format defined in RS-232. This pin creates an interface from the ZLF645 MCU to the serial port of a host PC using minimal external hardware. Figure 17 displays the recommended method of connecting P34 pin to an RS-232 connection using an open-drain buffer. The ICP pin must always be connected to V_{DD} through an external pull-up resistor.

For operation of the ICP, all power pins (V_{DD} *and* AV_{DD} *) must be supplied with power* and all ground pins (V_{SS} and AV_{SS}) must be properly grounded. **Caution:**

Figure 17. Interfacing the In-Circuit Programming Pin P34 with an RS-232 Interface (2)

ICP Data Format

The ICP interface uses the asynchronous data format defined for RS-232. Each character is transmitted as 1 Start bit, 8 data bits (least significant bit first), and 1.5 Stop bits (see Figure 18).

Figure 18. ICP Data Format

ICP Auto-Baud Detector/Generator

To run over a range of baud rates (data bits per second) with various system clock frequencies, the ICP contains an Auto-Baud Detector/Generator. After a reset, the ICP is non-active until it receives data. The ICP requires that the first character sent from the host is character 80H. The character 80H has eight continuous bits Low (one Start bit plus 7 data bits), framed between High bits. The Auto-Baud Detector measures this period and sets the ICP Baud Rate Generator accordingly.

The Auto-Baud Detector/Generator is clocked by the system clock. The minimum baud rate is the system clock frequency divided by 512. If the datastream can be synchronized with the system clock, the auto-baud generator can run as high as the system clock frequency divided by 2.

For optimal operation with asynchronous datastreams, the maximum recommended baud rate is the system clock frequency divided by 8. The maximum possible baud rate for asynchronous datastreams is the system clock frequency divided by 4, but this theoretical maximum is possible only for low noise designs with clean signals. Table 27 lists minimum and recommended maximum baud rates for sample crystal frequencies.

Table 27. ICP Baud-Rate Limits

If the ICP receives a Serial Break (nine or more continuous bits Low) the Auto-Baud Detector/Generator resets. You can reconfigure the Auto-Baud Detector/Generator by sending character 80H.

ICP Serial Errors

The ICP can detect any of the following error conditions on the P34 pin when in ICP mode:

- **•** Serial Break (a minimum of nine continuous bits Low).
- **•** Framing Error (received Stop bit is Low).
- **•** Transmit Collision (ICP and host simultaneous transmission detected by the ICP).

When the ICP detects one of these errors, it aborts any command currently in progress, transmits a four character long Serial Break back to the host, and resets the Auto-Baud Detector/Generator. A Framing Error or Transmit Collision can be caused by the host sending a Serial Break to the ICP. Because of the open-drain nature of the interface, returning a Serial Break break back to the host only extends the length of the Serial Break, if the host releases the Serial Break early.

The host transmits a Serial Break on the ICP pin when first connecting to the device or recovering from an error. A Serial Break from the host resets the Auto-Baud Generator/ Detector but does not resets the ICP Control Register. A Serial Break leaves the device in DEBUG mode if that is the current mode. The ICP is held in Reset until the end of the Serial Break when the ICP pin returns High. Because of the open-drain nature of the ICP pin, the host can send a Serial Break to the ICP even if the ICP is transmitting a character.

As the ICP interface uses a single pin for both receive and transmit, it can only receive or transmit at a given time. For the most part, this is not a problem, as the ICP uses a host driven protocol ($Z8^{\circ}$) does not send any data without the host asking for it).

To aid the ICP in avoiding collisions, the transmitter waits an additional 1/2 bit times after a Stop bit is fully received or transmitted before it starts transmission of a character. On the other hand, the receiver starts searching for a Start bit as soon as the middle of the Stop bit has been sampled and is valid. The transmitter does not start if another character is being received.

ICP In-Circuit Programming Commands

The host communicates to the ICP by sending ICP commands using the ICP interface. During normal operation, only a subset of the ICP commands are available. In FLASH CONTROL mode, all ICP commands are available, but for few commands their access to the Flash is qualified based upon the programming of the Flash Read/Write Protect Option bit (FLRWP) or the Lower Half Flash Read/Write Protect Option bit (FLPROT1). When either of these bits is enabled, some of the ICP commands will have reduced Flash memory access or will be disabled completely.

Table 28 is a summary of the ICP commands. Each ICP command is described in further detail in the bulleted list following this table. Table 28 also indicates those commands that operate when the device is not in FLASH CONTROL mode (normal operation) and how those commands are effected by programming of the FLRWP and FLPROT1 Option bits.

Table 28. In-Circuit Programmer Commands

Table 28. In-Circuit Programmer Commands (Continued)

In the following bulleted list of ICP commands, data and commands sent from the host to the ICP are identified by 'ICP \leftarrow Command/Data'. Data sent from the ICP back to the host is identified by ' $ICP \rightarrow Data$ ':

• Read ICP Revision (00H)—The Read ICP Revision command determines the version of the ICP. If ICP commands are added, removed, or changed, the revision number changes.

```
ICP \leftarrow 00HICP \rightarrow ICPREv[15:8] (Major revision number)
ICP \rightarrow ICPREv[7:0] (Minor revision number)
```


This command when executed returns a value of 0132H which is the revision ID assigned for the ZLF645 MCU.

• Read ICP Status Register (02H)—The Read ICP Status register command reads the ICPSTAT register.

```
ICP \leftarrow 02HICP \rightarrow ICPSTAT[7:0]
```
• Write ICP Control Register (04H)—The Write ICP Control register command writes the data that follows the command to the ICPCTL register.

```
ICP \leftarrow 04HICP \leftarrow ICPCTL[7:0]
```
• Read ICP Control Register (05H)—The Read ICP Control register command reads the value of the ICPCTL register.

```
ICP \leftarrow 05HICP \rightarrow ICPCTL[7:0]
```
• Write Flash Controller Registers (08H)—The Write Flash Controller register command allows writes to the Flash Controller registers. This command configures the Flash Controller for Flash memory accesses through the Write Flash Memory and Read Flash Memory commands. If the device is not in FLASH CONTROL mode, the register address and data values are discarded.

```
ICP \leftarrow 08HICP \leftarrow Register Address[15:0] ("OFH" for all Flash Ctrl Regs)
ICP \leftarrow \text{Request} Address[7:0]
ICP \leftarrow Size[7:0]ICP \leftarrow 1-256 data bytes
```
• Read Flash Controller Registers (09H)—The Read Flash Controller command allows reads of the Flash Controller registers. If the device is not in FLASH CONTROL mode this command returns FFH for all the register values.

```
ICP \leftarrow 09HICP \leftarrow Register Address[15:0] ("OFH" for all Flash Ctrl Regs)
ICP \leftarrow \text{Request} Address[7:0]
ICP \leftarrow Size[7:0]ICP \rightarrow 1-256 data bytes
```
• Write Flash Memory (0AH)—The Write Flash Memory command is used to write data to the main memory area or Information Area of the Flash memory. The command has equivalent functionality to the CPU writing the memory through the LDC and LDCI instructions. Data can be written 1 to memsize bytes at a time where memsize represents the size (32 KB or 64 KB) of the Flash memory for the product option chosen (The memsize number of bytes can be written by setting the size to 0). Should a size value greater than the maximum memory size be given by the user, the actual size value

for the command will default to the maximum memory size. The on-chip Flash Controller must be written to and unlocked for the programming operation to occur. If the Flash Controller is not unlocked, the data is discarded. Also, data is discarded for writes to protected areas of the Flash's main or information Page 3 areas based upon the settings of the read/write protect option bits in User Option Byte 1 (OPT1) register.

```
ICP \leftarrow 0AHICP \leftarrow Flash Memory Address[15:8]
ICP \leftarrow Flash Memory Address[7:0]
ICP \leftarrow Size[15:8]ICP \leftarrow Size[7:0]ICP \leftarrow 1-memsize data bytes
```
• Read Flash Memory (0BH)—The Read Flash Memory command is used to read data from the Flash's main memory area or Information Area. This command is equivalent to the CPU reading the memory through the LDC and LDCI instructions. Data can be read 1 to 'memsize' bytes at a time where memsize represents the size (32 KB or 64 KB) of the Flash memory for the product option chosen (The memsize number of bytes can be written by setting the size to 0). Depending on the settings of the read/write protect option bits in User Option Byte 1 register, reads to protected areas of the Flash's main memory area will return FFH for the data.

```
ICP \leftarrow 0BH
ICP \leftarrow Flash Memory Address[15:8]
ICP  Flash Memory Address[7:0]
ICP \leftarrow Size[15:8]ICP \leftarrow Size[7:0]ICP \rightarrow 1-65536 data bytes
```
• Read Flash Main Memory CRC (0EH)—The Read Flash Main Memory CRC command computes and returns the Cyclic Redundancy Check (CRC) of the Flash's Main Memory using the 16-bit CRC-CCITT polynomial. If the device is not in ICP mode, this command returns FFFFH for the CRC value. Unlike most other ICP Read commands, there is a delay from issuing of the command until the ICP returns the data. The ICP reads the Main Memory, calculates the CRC value, and returns the result. The delay is a function of the Flash main memory size and is approximately equal to the system clock period multiplied by the number of bytes in the Flash main memory.

```
ICP \leftarrow OEH
ICP \rightarrow CRC[15:8]ICP \rightarrow CRC[7:0]
```
• Read ICP Autobaud Register (1BH)— The Read ICP Autobaud register command reads the 12-bit ICP autobaud value set during autobaud detection.

```
ICP \leftarrow 1BHICP \rightarrow (4'b0000, \text{Autobaud}[11:8])
```


 $ICP \rightarrow \text{Autobaud}[7:0]$

• Write Test Mode Register (F0H)— The Write Test Mode Register command writes the data that follows the command to the TEST Mode Register (TESTMODE).

```
ICP \leftarrow F0HICP \leftarrow \text{TESTMODE}[7:0]
```
• Read Test Mode Register (F1H)— The Read Test Mode register command reads the value of the TESTMODE register.

```
ICP \leftarrow F1HICP \rightarrow TESTMODE[7:0]ICP \rightarrow \text{Autobaud}[7:0]
```
Flash Programming through the ICP Interface

Differences Between CPU Based and ICP Based Flash Programming/ Erase Access

Following are the differences for the allowed access capabilities between Flash accesses initiated by the CPU through instruction code and those initiated through the ICP interface:

- 1. The settings of the Flash Controller's Sector Protect Register (SPR) are ignored for Flash programming or page erase operations initiated through the ICP interface.
- 2. Mass erase operations can be executed through the ICP interface.

Using ICP Commands for Flash Programming/Read Operations

As described in the ICP In-Circuit Programming Commands, there are two commands that can be used for Flash programming and Flash data read operations. These commands are the **Write Flash Memory (0AH)** and **Read Flash Memory (0BH)** commands. To minimize the programming time required to program the Flash Memory using the ICP interface the following considerations concerning the use of these commands should be kept in mind:

• The **Write Flash Memory** command can be used in two different ways for transmitting Flash programming data. When 1 or more data bytes are to be programmed to one or more non-sequential Flash Memory address locations, a value of 0001H for the Size [15:8] and Size [7:0] arguments must be used. Using the command in this way requires that, for each byte of data to be programmed, 6 bytes be transmitted across the ICP. Following is an example of the ICP transmit sequence using the command, for programming two bytes of data:

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```
ICP \leftarrow 0AHICP \leftarrow Flash Memory Address1[15:8]
ICP \leftarrow Flash Memory Address1[7:0]
ICP \leftarrow 00HICP \leftarrow 01HICP \leftarrow Bytel[7:0]ICP \leftarrow 0AHICP \leftarrow Flash Memory Address2[15:8]
ICP \leftarrow Flash Memory Address2[7:0]
ICP \leftarrow 00HICP \leftarrow 01HICP \leftarrow \text{Byte2}[7:0]
```
If multiple bytes are to be programmed into sequential address locations in the Flash Memory, the **Write Flash Memory** command can be used so that each byte of data to be programmed only 1 byte be transmitted across the ICP, after the initial execution of the command. This is done simply by executing the command with a 'Size' value other than 0001H and providing the starting address of the Flash Memory area to be programmed. Following is an example of the ICP transmit sequence using the command, for programming 3 bytes of data to three sequential address locations of the Flash Memory:

```
ICP \leftarrow 0AHICP \leftarrow Starting Flash Memory Address[15:8]
ICP \leftarrow Starting Flash Memory Address[7:0]
ICP \leftarrow 00HICP \leftarrow 03HICP \leftarrow \text{Byte1}[7:0]ICP \leftarrow Byte2[7:0]
ICP \leftarrow Byte3[7:0]
```
• When using the **Write Flash Memory** command to program bytes of data into the Flash memory, there is no buffering of the data that takes place between the ICP interface and the Flash Memory. As a result the maximum rate at which data is programmed into the Flash Memory through the ICP interface is dependent up on how long it takes the ZLF645 to complete a Flash Memory byte programming operation, once it is initiated by the ICP. For the ZLF645, the total programming time required to program one byte of data is approximately 65 µs. When the **Write Flash Memory** command is used to program multiple bytes of data to sequential address locations in the Flash, then the maximum baud rate for Flash programming through the ICP is calculated as follows:

Max Programming Baud Rate = 1 ICP byte \times 10 ICP bits/byte/65 µs/byte = 153.8 kbaud

• If multiple non-sequential locations of the Flash Memory are to be programmed, the **Write Flash Memory** command can be still be used. However, as previously explained, each byte to be programmed requires 6 bytes be transmitted on the ICP interface. To keep the ICP interface data rate from limiting how quickly multiple bytes can be

programmed in this case a higher Baud rate can be used. Considering the ZLF645's system clock is of high frequency to support higher ICP Baud rates. The Baud rate necessary to support maximum programming efficiency is calculated as follows:

Max Baud Rate = 6 ICP byte \times 10 ICP bits/byte/65 μ s/byte = 922.8 kbaud

• The **Read Flash Memory** command can be used in the same two ways as described above for the **Write Flash Memory** command. When using the command to read multiple bytes of data from sequential address locations within the Flash memory, every byte read requires only 1 byte be received across the ICP interface. As described for the **Write Flash Memory**, there is no buffering of data that takes place between the ICP interface and the Flash Memory during memory reads. This means, as described for the **Write Flash Memory** command, the maximum Baud rate that memory read operations can occur at is dependent upon how quickly the ZLF645 completes a Flash Memory read operation, once it is initiated by the ICP. A ZLF645 memory read operation requires two system clock cycles to complete. Considering a ZLF645 system clock period of 250 ns, the theoretical maximum Baud rate reduces to the maximum Baud rate supported by the devices system clock frequency, which is calculated as follows:

Max Baud Rate = $1/(500 \text{ ns} / \text{bit}) = 2 \text{ M}$ baud

The ICP baud rate for read operations is significantly higher than for programming operations.

In-Circuit Programming Control Register Definitions

ICP Control Register

The ICP Control register (see Table 29) controls the state of the ICP interface. This register is used to enter or exit FLASH CONTROL mode.

Table 29. ICP Control Register (ICPCTL)

Bits			ю	◡		ω
Field	FLASHCTL			Reserved		
Reset						
R/W	R/W	r	R/W		D ↖	−

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ICP Status Register

The ICP Status register (see Table 30) reports status information about the current state of the ICP and the device.

Table 30. ICP Status Register (ICPSTAT)

TEST Mode Register

The TEST Mode register is used to enable various device test or Flash memory access modes. At present this register only provides configuration for a single mode where, once programmed, Flash memory accesses bypass the devices Flash Controller and are done through the devices I/O pins. A complete description of this mode is available in the Flash Byte Programming Interface section. This register can only be read or written using the ICP Read/Write Test Mode Register commands.

Table 31. TEST Mode Register (TESTMODE)

Exiting ICP Mode

The ZLF645 MCU is taken out of ICP mode under any of the following conditions:

- **•** Initiating a POR with P36 held High during the entire reset period.
- Lowering V_{DD} until the ZLF645 MCU reaches a Voltage Brownout reset state.

Flash Controller

Flash Memory Overview

The ZLF645 products feature either 32 KB or 64 KB of non-volatile Flash memory with read/write/erase capability. The Flash memory provides a 16-bit data interface but supports both 16-bit and 8-bit programming and read operations. The Flash memory can be programmed, read, or erased by the Flash Controller directed by either the CPU through user code or through the In-Circuit Programmer (ICP) interface pin with the ZLF645 in ICP mode. All user code or ICP Flash Accesses use the Flash's byte access mode where programs and reads occur 8 bits at a time. A Flash Byte Programming interface, as described in the section Flash Byte Programming Interface on page 82, is also available for Flash accesses through the devices GPIO pins and bypassing the Flash Controller. When the Flash Byte Programming interface is used, Flash programming and reads can be done either 8-bits or 16-bits at a time, depending on the package type of the device.

The Flash memory consists of two blocks, the **Main Memory** and the **Information Block**. The Flash main memory is arranged in pages with 512 bytes per page. Flash erasures are not allowed on a byte/word basis and a 512-byte page is the minimum Flash block size that can be erased. Each page is divided into 8 rows of 64 bytes.

The term 'page' in the context of the Flash Controller is not equivalent to the Z8® *LXMC CPU architecture's Program Memory page. For Flash contents protection, the Flash main memory is also divided into sectors, each sector containing 16 consecutive pages.* **Note:**

> In addition to the Flash main memory, there is a 256-byte Information block, arranged as 4 rows of 64 bytes. Each row is defined as a page. User access is only allowed to Page 3, where user definable Option bits reside. Pages 2-0 are for Maxim*®* internal use.

Information block does not have a Flash contents sector protection mechanism. **Note:**

> Table 32 lists the Flash main memory configuration for each device in the family of ZLF645 products. The size and configuration of the Information block is the same for all devices. Figure 19 displays the Flash memory arrangement.

Table 32. ZLF645 Products Flash Memory Configurations

Figure 19. Flash Memory Arrangement

Flash Information Block

The Flash Information Block of Flash memory is divided into two sections. Page 3 of the Information Block is accessible by you or Flash programmer vendor for programming, reading, or erasure through the ZLF645's ICP interface or it's Flash Byte Programming Interface only, as described in the section Flash Byte Programming Interface on page 82. The CPU has no access to this area of memory. User Option Bytes 0 and 1 use addresses 00FE and 00FF respectively of the Page 3 area and contain programmable bits with pre-defined functions.

The Flash read/write protect bits in User Option Byte 1 control the level of Page 3 access allowed to you along with the User's level of access to the Flash's main memory. Bytes 00C0 through 00FD of Page 3 have no pre-defined function and are available to you for other operations.

Pages 0 through 2 (addresses 0000 through 00BF), of the Information Area are reserved for Maxim[®] internal use and are inaccessible by you or programmer vendor, either through the ICP interface or by using the Flash Byte Programming interface.

Flash Controller Overview

The Flash Controller provides the appropriate Flash controls and timing for byte/word programming, Page Erase, Mass Erase, and reading of the Flash memory for Flash accesses made by either the CPU or through the ICP interface. It also limits programming, erase, and read access to the Flash memory based upon certain register and/or option bit settings. External accesses through the ZLF645's ICP or Flash Byte Programming Interfaces are limited by the Flash Controller based upon the programming of the ZLF645's Flash read/write protect bits in User Option Byte 1. Accesses by the CPU during code execution is limited based upon the programming of the Flash Controller's Sector Protect (FSEC) registers. All Flash memory accesses through the Flash Controller are prevented unless the Flash Controller is in 'unlocked' state.

Executing Flash Memory Accesses Through the Flash Controller

Flash Access Timing Control Programming Requirements

Before a program or erase operation can be executed by the Flash Controller on the Flash memory, you must first configure the Flash Controller's Flash frequency High and Low Byte registers. These registers combine to form a 16-bit value (FFREQ) that is used by the Flash Controller to control timing for Flash program and erase operations. For proper timing, the 16-bit binary Flash Frequency value must be programmed with the system clock frequency (in kHz).

This 16-bit binary Flash Frequency value is calculated using the following equation:

FFREQ[15:0] = System Clock Frequency (Hz)
1000

Using a 16-bit value FFREQ value, the Flash Controller is able to provide correct program and erase operation timing across a CPU clock frequency range of 1 MHz to 8 MHz.

 \sqrt{N}

Caution: The System Clock Frequency depends on the Flash memory programming of bit 2 of the *User Option Byte 1 and on the register programming of bit 0 of the SMR register and can be equal to the clock input frequency on the XTAL1 pin, a divide by 2 of that input, a divide by 16 of that input, or a divide by 32 of that input. Flash programming and erasure are not supported for CPU clock frequencies below 1 MHz or above 8 MHz. The Flash Frequency High and Low Byte registers must be loaded with the correct values.*

Enabling the Flash Controller For Flash Memory Accesses

Upon ZLF645 reset, the Flash Controller is put into a 'locked' state where Flash Accesses through the controller are disabled. Before any Flash memory accesses can take place through the Flash Controller it must be 'unlocked'. This functionality is designed to help protect against accidental programming or erasure of the Flash memory by Flash accesses initiated by the ICP interface or by the CPU during code execution. To 'unlock' the Flash Controller the ICP or CPU must perform the following sequence of Flash Controller Register write operations:

- 1. Program the Flash Controller's Page Select (PGS) register with the page to be programmed or erased.
- 2. Program the Flash Controller's Flash Control Register (FCTL) with a value of 73H.
- 3. Program the Flash Controller's Flash Control Register (FCTL) with a value of 8CH.
- 4. Program the Flash Controller's Page Select (PGS) register with the same value as programmed in step 1 above.

Failure to follow the exact register programming sequence as described above causes the Flash Controller to revert back to 'locked' state and the sequence must be repeated starting from step 1. For instance, if the two Page Select register writes in steps 1 and 4 do not match, the controller reverts to 'locked' state.

After 'unlocking' the Flash Controller, a programming or page erase operation can now be initiated through the Flash Controller to the page pointed to by the Page Select (PGS) register. For example, once the Flash Controller is 'unlocked', writing a 95H to the Flash Control (FCTL) register initiates a page erase. For a description of how to execute programming, see Byte Programming on page 74.

As mentioned in the Flash Memory Overview on page 67, CPU initiated programming or erase operations may be limited by the Flash Controller based upon the values programmed in the Flash Controller's Sector Protect (FSEC) register. For CPU initiated operations, the operation must be to a non-protected sector for the Flash Controller to execute the operation. For ICP initiated programming or erase operations, or if the page to be programmed/erased is in the Flash information Area, the operation is executed independent of the Sector Protect (FSEC) register settings.

After unlocking a specific page, the ICP or CPU can program any byte on that page or erase that page. For programming, after a byte is written the page remains unlocked, allowing for subsequent writes to other bytes on the same page. Once 'unlocked', the Flash Controller will revert to 'locked' state under the following conditions:

- 1. The Flash Controller has completed any programming operations in progress and the ICP or CPU writes the Flash Control (FCTL) register with a value other than 95H or 63H.
- 2. The Flash Controller has successfully completed a page erase or mass erase operation.

3. The CPU writes to the Page Select (PGS) register.

Figure 20 displays the basic Flash Controller operation considering code based CPU Flash accesses and based upon the programming of the Flash Controllers Flash Control (FCTL), Sector Protect (FSEC), and Page Select (FPS) Registers. As mentioned previously for ICP based Flash accesses, the only modification to Figure 20 is that the programming of the Sector Protect (FSEC) register is ignored and the ICP has programming and erase access to a page independent of whether it resides in a protected sector. Figure 20 does not display the effects of the Flash read/write protect bits of User Option byte 1.

If either of these bits is enabled, their function takes priority over the operation description displayed in Figure 20 in terms of when a page erase or byte programming access is allowed (for more details, see Flash Code Protection Against External Access on page 73).

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Figure 20. Flash Controller Operation Flow Chart

Flash Code Protection Against External Access

The Flash Controller limits Flash Access capabilities of the ICP and Flash Byte Programming Interfaces based upon the Flash read/write protect bits in User Option Byte 1. By programming these bits, you can configure the Flash Controller to block page 3 information area erasures, main memory reads, and main memory page erasures and programming as initiated through the ICP or Byte Programming Interfaces of the ZLF645. For more information, see Table 85 on page 174.

Flash Code Protection Against Accidental Program and Erasure

As mentioned previously, the ZLF645 products provide several levels of protection against accidental program and erasure of the Flash main memory contents by ICP and CPU accesses through the Flash Controller. Through the Flash Controller's register locking mechanism, page select redundancy, and sector level protection control, the ZLF645 products provide protection against accidental program and erasure of the Flash main memory contents by CPU and ICP accesses, except that for the ICP sector level protection is ignored. Similar levels of protection are in place for the Flash Information Area, minus the sector level protection.

Sector Based Flash Protection

For CPU initiated Flash main memory accesses, programming/erase protection is possible on a sector level basis through programming of the Flash Controller's Sector Protect (FSEC) register. For all ZLF645 products, each sector contains 16 pages (of 512 bytes each).

The Sector Protect (FSEC) register controls the protection state of each Flash sector. This register is address-shared with the Page Select register. It can only be accessed with the Flash Controller in 'locked' state. With the Flash Controller in 'locked' state, writing the Flash Control (FCTL) register with a value 5EH enables the Flash Controllers Sector Protect register to be written. The next write performed to Bank F, Register Address 02H then targets the Flash Controller's Sector Protect (FSEC) register.

The Sector Protect register is initialized to 0 on reset, putting each sector into an unprotected state. When a bit in the Sector Protect register is written to 1, the corresponding sector within the Flash memory can no longer be programmed or erased if for operations initiated by the CPU. Operations through the ICP are unaffected by the

settings of the Sector Protect (FSEC) register. After a bit of the Sector Protect register has been set, it cannot be cleared except by powering down the device.

Byte Programming

All Flash accesses either through CPU code execution or through the ICP interface occur using the Flash memory byte mode of operation. The Flash Controller allows CPU programming access to the Flash's main memory area only whereas the ICP has access to both the main memory and the page 3 information area for programming. The Flash memory is enabled for byte programming by either the CPU or the ICP after unlocking the Flash Controller and executing either a Mass Erase or Page Erase operation. When the Flash Controller is unlocked and a main memory Mass Erase is executed, all Flash Main Memory locations are available for byte programming by the CPU. In contrast, when the Flash Controller is unlocked and a main memroy Page Erase is executed, only the locations of the selected page as per the Page Select (PGS) register are available for byte programming by the CPU. An erased Flash byte contains all 1's (FFH).

The programming operation can only be used to change bits from 1 to 0. To change a Flash bit (or multiple bits) from 0 to 1 requires an erase operation through execution of either a Page Erase or Mass Erase command to the Flash Controller.

Byte Programming can be accomplished through the ICP by using the Write Memory command or by the Z8 LXMC CPU through execution of the LDC or LDCI instructions. For a description of the LDC and LDCI instructions, refer to *Z8*® *LXMC CPU User Manual (UM0215)*. During execution of a CPU initiated programming operation the system clock to the CPU is halted preventing further code execution, however the system clock and the on-chip peripherals continue to operate. Once the programming operation is complete, the CPU resumes code execution. To exit programming mode and lock the Flash the CPU can perform a write of any value to the Flash Control (FCTL) register, except the Mass Erase or Page Erase commands.

Page Erase

The Flash main memory can be erased one page (512 bytes) at a time. Page Erasing the Flash memory sets all bytes in that page to the value FFH. The Flash Page Select (PGS) register identifies the page to be erased. For CPU initiated page erase operations, only a page residing in an unprotected sector can be erased. With the Flash Controller unlocked and the active page set, writing the value 95H to the Flash Control (FCTL) register initiates the Page Erase operation. As with programming, during execution of a CPU initiated page erase operation the system clock to the CPU is halted preventing further code execution, however the system clock and the on-chip peripherals continue to operate. Once the page erase operation is complete, the CPU resumes code execution.

If a Page Erase operation to the Flash's main memory is performed using the ICP, bit 3 of the ICP Status register can be polled to determine when the operation is complete. When the Page Erase is complete, the Flash Controller returns to its locked state. Although the

Flash Controller prevents CPU accesses to the Flash's Information block, the ICP can initiate a Page erase to page 3 of Information Area by a similar process as used for the main memory. The only difference is that the ICP must first write bit 7 of the Flash Page Select (PGS) register to a 1 before writing the page erase command to the Flash Control (FCTL) register. For more details, see Table 34 on page 77.

Mass Erase

The Flash main memory can also be Mass Erased using the Flash Controller, but only through the ICP interface and not by the CPU. Mass Erasing the Flash memory sets all bytes to the value FFH. With the Flash Controller unlocked, writing the value 63H to the Flash Control register initiates the Mass Erase operation. If a Mass Erase operation is performed using the ICP, bit 3 of the ICP Status register is polled to determine when the operation is complete. When the Mass Erase is complete, the Flash Controller returns to its locked state. You cannot mass erase the Information Area.

If either of the Flash Memory Protect Option Bits are set as defined in the Flash Option Bits section, a mass erase of the Flash's main memory must be performed before Page 3 of the Flash's Information Area can be erased. These two operations must be done when the device is at operating voltage. That is, if a mass erase is followed with a power-down then power-up sequence, performing an Information Area Page 3 erase will not erase its contents. **Caution:**

Flash Control Register Definitions

Flash Control Register

The Flash Controller must be unlocked using the Flash Control (FCTL) register (see Table 33) before the Flash Controller is enabled for programming or erasing the Flash memory. Writing values of 73H and then 8CH sequentially to the Flash Control register unlocks the Flash Controller, as long as the other conditions described in Enabling the Flash Controller For Flash Memory Accesses on page 70 have been met. When the Flash Controller is unlocked, a Mass Erase initiated by the ICP, or Page Erase initiated by the ICP or CPU can be executed by the Flash Controller by writing the appropriate command value to this register. Execution of a Page Erase applies only to the active page selected in Flash Page Select (FPS) register. Writing an invalid value or an invalid sequence returns the Flash Controller to its locked state. The Write-only Flash Control register shares its Register File address with the Read-only Flash Status register.

Table 33. Flash Control Register (FCTL)

Flash Status Register

The Flash Status (FSTAT) register (see Table 34) indicates the current state of the Flash Controller. This register can be read any time. The read-only Flash Status (FSTAT) register shares its Register File address with the Write-only Flash Control (FCTL) register.

Table 34. Flash Status Register (FSTAT)

Flash Page Select Register

The Flash Page Select (FPS) register (see Table 35) shares address space with the Flash Sector Protect (FSEC) register. Unless the Flash Controller is in 'locked' state and its Flash Control (FCTL) register is written with 5EH, writes to this address target the Flash Page Select (FPS) register.

The FPS register is used to select one page within the Flash Main Memory or Information Block for programming or erasure depending upon whether its IFEN bit is 0 or 1 respectively. Each Flash Main Memory Page contains 512 bytes of Flash memory. During a Page Erase operation to the Main Memory, the page that will be erased is the one containing the 512 Flash memory locations where bits 15 through 9 of their addresses is equal to bits 6 through 0 of FPS register. For Main Memory programming operations, bits 15 through 9 of the address to be programmed must equal bits 6 through 0 of the FPS register for the Flash Controller to execute the operation. For page erase or programming operations to the Flash's Information Block as indicated by the IFEN bit being 1, the programming or

page erase command must be initiated by the ICP. Information Block page erase or programming operations initiated by the CPU are ignored by the Flash Controller. In the case of an Information Block programming or page erase operation initiated by the ICP, the FPS register must first be programmed with 83H to point to page 3 of the Information Block or else the operation will be ignored by the Flash Controller. For Information Block programming through the ICP, bits 12 through 6 of the address must equal bits 6 through 0 of the FPS register for the Flash Controller to execute the operation.

Table 35. Flash Page Select Register (FPS)

Bits		6	5	4	3			
Field	IFEN	PAGE						
Reset		0	0				0	
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Address	Bank F, Register address: 02H							

Flash Sector Protect Register

The Flash Sector Protect (FSEC) register (see Table 36) address is shared with the Flash Page Select (FPS) register. It is accessed by first writing 5EH to the Flash Control (FCTL) register with the Flash Controller in 'locked' state, and then writing to the register file address location given for the Flash Page Select (FPS) register.

The FSEC register selects which of the eight available Flash memory sectors is to be protected from CPU initiated programming or page erase operations. For ICP initiated programming or page erase operations, the settings within the FSEC register are ignored by the Flash Controller. The reset state of each Sector Protect bit in the FSEC register is its unprotected state or 0 value. After a sector is protected by setting its corresponding register bit to 1, it cannot be unprotected (the register bit cannot be cleared) without powering down the device.

Table 36. Flash Sector Protect Register (FSEC)

Flash Frequency High and Low Byte Registers

The Flash Frequency High and Low Byte registers (Table 37 and Table 38) combine to form a 16-bit value, FFREQ, to control timing for Flash program and erase operations. The 16-bit binary Flash frequency value must contain the system clock frequency (in kHz) and is calculated using the following equation:

```
FFREQ[15:0] = {FFREQH[7:0], FFREQL[7:0]} = \frac{\textrm{System Clock Frequency (Hz)}}{1000}
```
Programming the Flash Frequency High and Low Byte Registers as per the formula provides a Flash programming time of approximately $30 \mu s$ and an erase time of approximately 10 ms.

Flash programming and erasure is not supported for system clock frequencies below 1 MHz or above 8 MHz. The Flash Frequency High and Low Byte registers must be loaded with the correct value to ensure proper operation of the device.

Table 37. Flash Frequency High Byte Register (FFREQH)

Table 38. Flash Frequency Low Byte Register (FFREQL)

Flash Controller Functions Summary

The Flash Controller performs its functions, directed by either the ICP interface or by the CPU through instruction codes. Table 39 lists the functions that will or will not be performed by the Flash Controller, according to whether the CPU or ICP is the initiator, whether the operation is performed on the Flash's Main Memory or Information Block, and whether either of the two read/write protect bits of User Option Byte 1 have been enabled or not.

Table 39. Flash Controller Functions Summary

Notes

1. FLPROT1 = 0, cannot read or write lowest half of memory. FLRWP = 0, cannot read or write entire main memory.

2. FLRWP/FLPROT1 = 0, cannot write or erase Page 3.

3. FLPROT1 = 0 , no effect. FLRWP = 0 , no effect.

Flash Byte Programming Interface

Using the ZLF645's Flash Byte Programming interface, the on-chip Flash controller can be bypassed, allowing direct control of the Flash signals through registered values of certain of the ZLF645's GPIO pins. Bypassing the Flash controller allows faster row programming algorithms to be used by controlling the Flash programming signals directly. This method is beneficial when programming a large number of devices and can be used for Flash programming by third party vendors who manufacture gang programmers. For more information on how to use this interface, refer to *Third-Party Flash Programming Support for Z8 Crimzon Flash Parts*, available for download at www.maxim-ic.com.

Enabling The Flash Byte Programming Interface

The Flash Byte Programming Interface is enabled by writing three bytes to the ICP interface:

- 1. 80H initiates auto-baud calculation of the ICP interface data and clock rate.
- 2. F0H ICP Write Test Mode Register command.
- 3. 04H Data to be written to the Test Mode Register. This enables the Flash Byte Programming interface.

Since Flash Byte Programming Interface is enabled with the ZLF645 MCU in ICP mode, the CPU clock will stop and no CPU accesses to the Flash memory will occur. **Note:**

Flash Byte Programming Interface Flash Access Restrictions

The types of Flash access allowed to the Flash memory through the Flash Byte Programming interface is qualified similar to the ICP, by the settings of the Flash Memory Protection Bits in User Option Byte 1. If either of the Flash protect bits are set, the program memory has to be mass erased before full read/program access is allowed to either the main memory or Information area page 3 sections of the Flash memory, respectively. Flash memory access allowed through the Flash Byte Programming interface is summarized in Table 40.

Table 40.Flash Byte Programming Functions Summary

Notes

1. Program, Read, and Page Erase access is limited to the upper half address space of the main memory only.

2. Only Page 3 of the Information Area is accessible for Program, Read, and Page Erase operations.

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Infrared Learning Amplifier

The ZLF645 MCU's infrared learning amplifier allows you to detect and decode infrared transmissions directly from the output of the receiving diode without the need for external circuitry (see Port 3 on page 23).

An IR diode can be connected to the IR amplifier as displayed in Figure 21. When the IR amplifier is enabled and an input current is detected on Port 3, Pin 1 (P31), the IR amplifier outputs a logical High value. When the input current is below the switching threshold of the IR amplifier, the amplifier outputs a logical Low value.

Within the MCU, the IR amp output goes to the capture/timer logic, which can be programmed to demodulate the IR signal. The IR amplifier output can also be read by the CPU, or drive the Port 3, Pin 4 (P34) output if write-only register bit PCON[0] is set to 1.

For the maximum current input that is clearly recognized by the ZLF645 as a 0 and the minimum current input that is clearly recognized as a 1, see I_{DETHO} and I_{DETHI} parameters, respectively, in Table 80 on page 165.

The IR learning amplifier can demodulate signals up to a frequency of 500 kHz. A special mode exists that allows you to capture the third, fourth, and fifth edges of the IR amplifier output and generate an interrupt.

Figure 21. Learning Amplification Circuitry within the ZLF645 Flash MCU

For details on programming the timers to demodulate a received signal, see Timers on page 99.

Universal Asynchronous Receiver/ Transmitter

The Universal Asynchronous Receiver/Transmitter (UART) is a full-duplex communication channel capable of handling asynchronous data transfers. The two UARTs use a single 8-bit data mode with selectable parity.

The UART interface when enabled uses the GPIO pins P07 for the UART transmit and P32 for the UART receive.

The features of the UART include:

- **•** 8-bit asynchronous data transfer
- Selectable even- and odd-parity generation and checking
- One or two Stop bits
- Separate transmit and receive interrupts
- Framing, overrun, and break detection
- **•** Separate transmit and receive enables
- **•** 8-bit Baud Rate Generator
- **•** Baud Rate Generator timer mode
- **•** UART operational during HALT mode

Table 41. UART Control Registers

Address (Hex)

Architecture

The UARTs consist of three primary functional blocks: **transmitter**, **receiver**, and **Baud Rate Generator**. The UART transmitter and receiver function independently, but employ the same baud rate and data format. Figure 22 displays the UART architecture.

Figure 22. UART Block Diagram

Operation

The UART channel can be used to communicate with a master microprocessor or a slave microprocessor, both of which exhibit transmit and receive functionality. You can either operate the UART channel by polling the UART Status register or via interrupts. The UART remains active during HALT mode. If neither the transmitter nor the receiver is enabled, the UART baud rate generator can be used as an additional timer. The UART contains a noise filter for the receiver that can be enabled by the user.

Data Format

The UART transmits and receives data in an 8-bit data format, with the least significant bit (lsb) occurring first. An even- or odd-parity bit can be optionally added to the data stream. Each character begins with an active Low Start bit and ends with either 1 or 2 active High Stop bits. Figure 23 and Figure 24 display the asynchronous data format employed by the UARTs with or without parity, respectively.

Figure 23. UART Asynchronous Data Format without Parity

Figure 24. UART Asynchronous Data Format with Parity

Transmitting Data Using Polled Method

Follow the steps below to transmit data using the polled method of operation:

- 1. Write to the Baud Rate Generator Constant (BCNST) register, address 0F4h, to set the appropriate baud rate.
- 2. Write 0 to bit 6 of the P01M register.
- 3. Write to the UART Control register (UCTL) to:
	- (a) Set the transmit enable bit, UCTL[7], to enable the UART for data transmission.
	- (b) If parity is appropriate, set the parity enable bit, UCTL[4] to 1 and select either even- or odd-parity (UCTL[3]).

- 4. Check the Transmit Status register bit, UST[2], to determine if the Transmit Data register is empty (indicated by 1). If empty, continue to Step 6. If the Transmit Data register is full (indicated by 0), continue to monitor the UST[2] bit until the Transmit Data register is available to receive new data.
- 5. Write the data byte to the UART Transmit Data register, 0F1h. The transmitter automatically transfers the data to the internal transmit shift register and transmits the data.
- 6. To transmit additional bytes, return to Step 4.
- 7. Before disabling the transmitter, read the transmit completion status bit, UST[1]. If $\text{UST}[1]=0$, continue to monitor the bit until it changes to 1, which indicates that all data in the Transmit Data and internal shift registers has been transmitted.

Data written while the transmit enable bit is clear (UCTL[7]=0) will not be transmitted. Data written while the transmit data status bit is clear (UST[2]=0) overwrites the previous value written, so the previous written value will not be transmitted. Disabling the UART transmitter while the transmit completion status bit is clear (UST[1]=0) can corrupt the byte being transmitted.

Transmitting Data Using Interrupt-Driven Method

The UART transmitter interrupt indicates the availability of the Transmit Data register to accept new data for transmission.

Follow the steps below to configure the UART for interrupt-driven data transmission:

- 1. Write to the BCNST register to set the appropriate baud rate.
- 2. Write 0 to bit 6 of the P01M register.
- 3. Execute DI instruction to disable interrupts.
- 4. Write to the Interrupt Control registers to enable the UART Transmitter interrupt and set the appropriate priority.
- 5. Write to the UART Control register to:
	- (a) Set the transmit enable bit (UCTL bit 7) to enable the UART for data transmission.
	- (b) Enable parity, if appropriate, and select either even- or odd-parity.
- 6. Execute an EI instruction to enable interrupts as the transmit buffer is empty, an interrupt is immediately executed.
- 7. Write the data byte to the UART Transmit Data register. The transmitter automatically transfers the data to the Internal Transmit Shift register and transmits the data.
- 8. Execute the IRET instruction to return from the interrupt service routine (ISR) and wait for the Transmit Data register to again become empty.

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- 9. Before disabling the transmitter, read the transmit completion status bit, UST[1]. If $\text{UST}[1]=0$, continue to monitor the bit until it changes to 1, which indicates that all data in the Transmit Data and Internal Shift registers has been transmitted.
- **Caution:**
- *Data written while the transmit enable bit is clear (UCTL[7]=0) will not be transmitted. Data written while the transmit data status bit is clear (UST[2]=0) overwrites the previous value written, so the previous written value will not be transmitted. Disabling the UART transmitter while the transmit completion status bit is clear (UST[1]=0) can corrupt the byte being transmitted.*

Receiving Data Using the Polled Method

Follow the steps below to configure the UART for polled data reception:

- 1. Write to the BCNST register to set the appropriate baud rate.
- 2. Write to the UART Control register (UCTL) to:
	- (a) Set the receive enable bit (UCTL[6]) to enable the UART for data reception
	- (b) Enable parity (if appropriate) and select either even- or odd-parity
- 3. Check the receive status bit in the UART Status register, bit UST[7], to determine if the Receive Data register contains a valid data byte (indicated by a 1). If UST[7] is set to 1 to indicate available data, continue to $Step 4$. If the Receive Data register is empty (indicated by a 0), continue to monitor the UST[7] bit awaiting reception of the valid data.
- 4. Read data from the UART Receive Data register.
- 5. Return to Step 3 to receive additional data.

Receiving Data Using the Interrupt-Driven Method

The UART Receiver interrupt indicates the availability of new data (as well as error conditions).

Follow the steps below to configure the UART receiver for interrupt-driven operation:

- 1. Write to the UART BRG Constant registers to set the appropriate baud rate.
- 2. Execute DI instruction to disable interrupts.
- 3. Write to the Interrupt Control registers to enable the UART receiver interrupt and set the appropriate priority.
- 4. Clear the UART Receiver interrupt in the applicable Interrupt Request register.
- 5. Write to the UART Control register (UCTL) to:
	- (a) Set the receive enable bit (UCTL[6]) to enable the UART for data reception
	- (b) Enable parity, if appropriate, and select either even- or odd-parity
- 6. Execute an EI instruction to enable interrupts.

The UART is now configured for interrupt-driven data reception. When the UART Receiver interrupt is detected, the associated ISR performs the following:

- 1. Checks the UART Status register to determine the source of the interrupt, whether it is an error, break, or received data.
- 2. Reads the data from the UART Receive Data register, if the interrupt was caused by data available.
- 3. Clears the UART receiver interrupt in the applicable Interrupt Request register.
- 4. Executes the IRET instruction to return from the ISR and await more data.

UART Interrupts

The UART features separate interrupts for the transmitter and the receiver. In addition, when the UART primary functionality is disabled, the BRG can also function as a basic timer with interrupt capability.

When the UART is set to run at higher baud rates, the UART receiver's service routine may not have enough time to read and manipulate all bits in the UART Status register (especially bits generating error conditions) for a received byte before the next byte is received. You can devise your own hand-shaking protocol to prevent the transmitter from transmitting more data while current data is being serviced. **Note:**

Transmitter Interrupts

The transmitter generates a single interrupt when the Transmit Status bit, UST[2], is set to 1. This indicates that the transmitter is ready to accept new data for transmission. The Transmit Status interrupt occurs after the internal transmit shift register has shifted the first bit of data out. At this point, the Transmit Data register can be written with the next character to send. This provides 7 bit periods of latency to load the Transmit Data register before the transmit shift register completes shifting the current character. Writing to the UART Transmit Data register clears the UST[2] bit to 0. The interrupt is cleared by writing a 0 to the Transmit Data register.

Receiver Interrupts

The receiver generates an interrupt when any of the following occurs:

- **• A data byte is received and available in the UART Receive Data register—**This interrupt can be disabled independent of the other receiver interrupt sources. The received data interrupt occurs once the receive character has been received and placed in the Receive Data register. Software must respond to this received data available condition before the next character is completely received to avoid an overrun error. The interrupt is cleared by reading from the UART Receive Data register.
- **• A break is received**—A break is detected when a 0 is sent to the receiver for the full byte plus the parity and stop bits. After a break is detected, it will interrupt

immediately if there is no valid data in the Receive Data register. If data is present in the Receive Data register, an interrupt will occur after the UART Receive Data register is read.

- **• An overrun is detected—**An overrun occurs when a byte of data is received while there is valid data in the UART Receive Data register that has not been read by the user. The interrupt will be generated when the user reads the UART Receive Data register. The interrupt is cleared by reading the UART Receive Data register. When an overrun error occurs, the additional data byte will not overwrite the data currently stored in the UART Receive Data register.
- **• A data framing error is detected**—A data framing error is detected when the first stop bit is 0 instead of 1. When configured for 2 stop bits, a data framing error is only detected when the first stop bit is 0. A framing error interrupt is generated when the framing error is detected. Reading the UART Receive Data register clears the interrupt.

Ensure that the transmitter uses the same stop bit configuration as the receiver. **Note:**

UART Overrun Errors

When an overrun error condition occurs the UART prevents overwriting of the valid data currently in the Receive Data register. The Break Detect and Overrun status bits are not displayed until after the valid data has been read.

After the valid data has been read, the UART Status (UST) register is updated to indicate the overrun condition (and Break Detect, if applicable). The UST[7] bit is set to 1 to indicate that the Receive Data register contains a data byte. However, because the overrun error occurred, this byte may not contain valid data and must be ignored. The Break Detect bit, UST[3], indicates if the overrun was caused by a break condition on the line. After reading the status byte indicating an overrun error, the Receive Data register must be read again to clear the error bits is the UART Status 0 register. Updates to the Receive Data register occur only when the next data word is received.

UART Data and Error Handling Procedure

Figure 25 on page 92 displays the recommended procedure for use in UART receiver interrupt service routine.

Figure 25. UART Receiver Interrupt Service Routine Flow

Baud Rate Generator Interrupts

If the BRG interrupt enable is set, the UART Receiver interrupt asserts when the UART Baud Rate Generator reloads. This action allows the BRG to function as an additional counter if the UART functionality is not employed.

UART Baud Rate Generator

The UART Baud Rate Generator creates a lower frequency baud rate clock for data transmission. The input to the Baud Rate Generator is the system clock. The UART Baud Rate Constant register contains an 8-bit baud rate divisor value (BCNST[7:0]) that sets the data transmission rate (baud rate) of the UART. For programmed register values other than 00h, the UART data rate is calculated using the below equation:

699.
$$
-
$$

\n699. $-$

\n699. $-$

\n716 x UART Baud Rate Division Value (B)

When the UART Baud Rate Low register is programmed to 00h, the UART data rate is calculated as follows:

UART Data Rate (bps) = $\frac{\text{System Clock Frequency (Hz)}}{\text{System}}$ 4096

When the UART Baud Rate Generator is used as a general-purpose counter, the counters time-out period can be computed as follows based upon the counters clock input being a divide by 16 of the system clock and the maximum count value being 255:

Time-Out Period (us) $=$ 16 x UART Baud Rate Divisor Value (BCNST) System Clock Frequency (MHz)

The relationship between the XTAL1 clock frequency and the system clock frequency must be considered before making this computation and is dependent upon the programming of bit 2 of User Option Byte 1 as well as the programming of bit 0 of the SMR register. Depending on the programmed values, the system clock frequency can be a divide by 1, a divide by 2, or a divide by 16 of the XTAL1 clock. **Note:**

> When the UART is disabled, the BRG can function as a basic 8-bit timer with interrupt on time-out.

Follow the steps below to configure the BRG as a timer with interrupt on time-out:

- 1. Disable the UART by clearing the receive and transmit enable bits, UCTL[7:6] to 0.
- 2. Load the appropriate 8-bit count value into the UART Baud Rate Generator Constant register. The count frequency is the system clock frequency (in Hz) divided by 16.
- 3. Enable the Baud Rate Generator timer function and associated interrupt by setting the Baud Rate Generator bit (UCTL bit 0) in the UART Control register to 1. When configured as an 8-bit timer, the count value, instead of the reload value, is read, and the counter begins counting down from its initial programmed value. On timing out (reaching a value of 1), if the time-out interrupt is enabled, an interrupt will be produced. The counter will then reload its programmed start value and begin counting down again.

Table 42 lists a number of BCNST register settings at various baud rates and system clock frequencies.

Table 42. BCNST Register Settings Examples

UART Receive Data Register/UART Transmit Data Register

The UART Receive/Transmit Data register (see Table 43) is used to send and retrieve data from the UART channel. When the UART receives a data byte, it can be read from this register. The UART receive interrupt is cleared when this register is used. Data written to this register is transmitted by the UART.

Table 43. UART Receive/Transmit Data Register (URDATA/UTDATA)

UART Status Register

The UART Status register (see Table 44) displays the status of the UART. Bits [6:3] are cleared by reading the UART Receive/Transmit register (F1h).

Table 44. UART Status Register (UST)

UART Control Register

The UART Control register controls the UART. In addition to setting bit 5, you must also set appropriate bit in the Interrupt Mask register (see Table 65 on page 133).

This register is not reset after a Stop Mode Recovery. **Note:**

▶

Table 45. UART Control Register (UCTL)

UART Baud Rate Generator Constant Register

The UART baud rate generator determines the frequency at which UART data is received and transmitted. This baud rate is determined by the following equation:

UART Data Rate (bps) = $\frac{System Clock Frequency (Hz)}{|E|}$ 16 x UART Baud Rate Divisor Value (BCNST)

The system clock is usually the crystal clock divided by 2. When the UART baud rate generator is used as an additional timer, a Read from this register returns the actual value of the count of the BRG in progress and not the reload value. See Table 46.

This register is not reset after a Stop Mode Recovery. **Note:**

Table 46. UART Baud Rate Generator Constant Register (BCNST)

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Timers

The ZLF645 MCU infrared timer features a 16-bit and an 8-bit counter/timer, each of which can be used simultaneously for transmitting. Both timers can be used for demodulating an input carrier wave and share a single input pin.

Figure 26 displays the counter/timer architecture, which is designed to help unburden the program from coping with real-time problems like generating complex waveforms or receiving and demodulating complex waveforms and pulses.

In addition to the 16-bit and 8-bit timers, the UART's baud rate generator can be used as an additional 8-bit timer when the UART receiver is not in use (for more details, see Universal Asynchronous Receiver/Transmitter on page 85).

Figure 26. Counter/Timers Block Diagram

Table 47 summarizes the timer control registers. Some timer functions can also be affected by control registers for other peripheral functions.

Address (Hex)

Counter/Timer Functional Blocks

The ZLF645 MCU infrared timer contains a glitch filter for removing noise from the input when demodulating an input carrier. Each timer features its own demodulating mode and can be simultaneously used to generate a signal output. The T8 timer has the ability to capture only one cycle of a carrier wave of a high-frequency waveform.

Input Circuit

Depending on the setting of register bits P3M[2:1] and CTR1[6], the timer/counter input circuit monitors one of the following conditions:

- **•** The P31 digital signal, if CTR1[6]=0 and P3M[2:1]=00.
- The P31 analog comparator output, if CTR1[6]=0 and P3M[2:1]=01.
- **•** The P31 IR amplifier output, if CTR1[6]=0 and P3M[2]=1.
- The P20 digital signal, if CTR1[6]=1.

Based on register bits CTR1[5:4], a pulse is generated at when a rising edge, falling edge, or any edge is detected. Glitches in the input signal are filtered out if they are shorter than the glitch filter width specified in register bits CTR1[3:2]. Figure 27 displays the input circuit.

Figure 27. Counter/Timer Input Circuit

The timers can be configured to operate in following modes:

- **•** T8 TRANSMIT Mode
- **•** T8 DEMODULATION Mode
- **•** T16 TRANSMIT Mode
- **•** T16 DEMODULATION Mode
- **•** PING-PONG Mode

T8 TRANSMIT Mode

Before T8 is enabled, the output of T8 depends on CTR1, bit 1. If the bit is 0, T8_OUT is 1; if it is 1, T8_OUT is 0. See Figure 28.

Figure 28. TRANSMIT Mode Flowchart

When T8 is enabled, the output T8_OUT switches to the initial value (CTR1, bit 1). If the initial value (CTR1, bit 1) is 0, TC8L is loaded; otherwise, TC8H is loaded into the counter.

In SINGLE-PASS mode (CTR0, bit 6), T8 counts down to 0 and stops, T8_OUT toggles, the time-out status bit (CTR0, bit 5) is set, and a time-out interrupt can be generated if it is enabled (CTR0, bit 1).

In MODULO-N mode, on reaching terminal count, T8_OUT is toggled, but no interrupt is generated. From that point, T8 loads a new count (if the T8_OUT level now is 0), TC8L is loaded; if it is 1, TC8H is loaded. T8 counts down to 0, toggles T8_OUT, and sets the time-out status bit (CTR0, bit 5), thereby generating an interrupt if enabled (CTR0, bit 1). One cycle is thus completed. T8 then loads from TC8H or TC8L according to the T8_OUT level and repeats the cycle. See Figure 29.

Figure 29. 8-Bit Counter/Timer Circuits

You can modify the values in TC8H or TC8L at any time. The new values take effect when they are loaded.

An initial count of 1 is not allowed (a non-function occurs). An initial count of 0 causes TC8 **Caution:***to count from 0 to FFh to FEh.*

/I\

1. *The "h" suffix denotes hexadecimal values.* **Notes:**

2. *Transition from 0 to FFh is not a time-out condition.*

Using the same instructions for stopping the counter/timers and setting the status bits is not **Caution:** *recommended.*

> Two successive commands are necessary. First, the counter/timers must be stopped. Second, the status bits must be reset. These commands are required as it takes one counter/ timer clock interval for the initiated event to actually occur. See Figure 30 and Figure 31.

T8 DEMODULATION Mode

You must program TC8L and TC8H to FFh. After T8 is enabled, when the first edge (rising, falling, or both depending on CTR1 bits [5:4]) is detected, it starts to count down. When a subsequent edge (rising, falling, or both depending on CTR1 bits [5:4]) is detected during counting, the current value of T8 is complemented and put into one of the capture registers.

If it is a positive edge, data is put into LO8; if it is a negative edge, data is put into HI8. From that point, one of the edge detect status bits (CTR1, bits [1:0]) is set, and an interrupt can be generated if enabled (CTR0, bit 2). Meanwhile, T8 is loaded with FFh and starts counting again. If T8 reaches 0, the time-out status bit (CTR0, bit 5) is set, and an interrupt can be generated if enabled (CTR0, bit 1). T8 then continues counting from FFh(see Figure 32 on page 106).

When bit 4 of CTR3 is enabled, the flow of the demodulation sequence is altered. The third edge makes T8 active, and the fourth and fifth edges are captured. The capture interrupt is activated after the fifth event occurs. This mode is useful for capturing the carrier duty cycle as well as the frequency at which the first cycle is corrupted. See Figure 33 and Figure 34.

Figure 33. DEMODULATION Mode Flowchart

Figure 34. DEMODULATION Mode Flowchart with Bit 4 of CTR3 Set

T16 TRANSMIT Mode

In NORMAL or PING-PONG mode, the output of T16 when not enabled depends on CTR1, bit 0. If this bit is set to 0, T16 OUT is a 1; if set to 1, T16 OUT is 0. You can force the output of T16 to either 0 or 1 whether it is enabled or not by programming CTR1 bits [3:2] to a 10 or 11.

When bit 4 of CTR3 is set, the T16 output does not update. However, time-out interrupts (flags) are still updated. In addition, the T8 carrier is not disrupted by timing out of the T16 timer.

When T16 is enabled, a value of $(TC16H * 256) + TC16L$ is loaded, and T16 OUT is switched to its initial value (CTR1, bit 0). When T16 counts down to 0, T16 OUT is toggled (in NORMAL or PING-PONG mode), an interrupt (CTR2, bit 1) is generated (if enabled), and a status bit (CTR2, bit 5) is set. See Figure 35.

Figure 35. 16-Bit Counter/Timer Circuits

Global interrupts override this function as described in the Interrupts on page 127. **Note:**

> If T16 is in SINGLE-PASS mode, it is stopped at this point (see Figure 36 on page 110). If it is in MODULO-N mode, it is loaded with $TC16H * 256 + TC16L$, and the counting continues (see Figure 37 on page 110). You can modify the values in TC16H and TC16L at any time. The new values take effect when they are loaded.

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Do not load these registers at the time the values are to be loaded into the counter/timer to **Caution:** \sqrt{N} *ensure known operation. An initial count of 1 is not allowed. An initial count of 0 causes T16 to count from 0 to* FFFEh*. Transition from 0 to* FFFFh *is not a time-out condition.*

Counter Enable command; T16_OUT, switches to its initial value (CTR1 data bit 0)

Time-out interrupt Time-out interrupt T16_OUT toggles, T16_OUT toggles,

Figure 37. T16_OUT in MODULO-N Mode

T16 DEMODULATION Mode

You must program TC16L and TC16H to FFh. After T16 is enabled, and the first edge (rising, falling, or both depending on CTR1 bits [5:4]) is detected, T16 captures HI16 and LO16, reloads, and begins counting.

If Bit 6 of CTR2 Is 0—When a subsequent edge (rising, falling, or both depending on CTR1 bits [5:4]) is detected during counting, the current count in T16 is complemented and loaded into HI16 and LO16. When data is captured, one of the edge detect status bits (CTR1, bit 1; bit 0) is set, and an interrupt is generated if enabled (CTR2, Bit 2). T16 is loaded with FFFFh and starts again. This T16 mode is generally used to measure space time, the length of time between bursts of carrier signal (marks).

If Bit 6 of CTR2 Is 1—T16 ignores the subsequent edges in the input signal and continues counting down. A time-out of T8 causes T16 to capture its current value and generate an interrupt if enabled (CTR2, Bit 2). In this case, T16 does not reload and continues counting. If CTR2 bit 6 is toggled (by writing a 0 then a 1 to it), T16 captures and reloads on the next edge (rising, falling, or both depending on CTR1 bits [5:4]), continuing to ignore subsequent edges. This T16 mode generally measures mark time, the length of an active carrier signal burst.

If T16 reaches 0, T16 continues counting from FFFFh. Meanwhile, a status bit (CTR2 bit 5) is set, and an interrupt time-out can be generated, if enabled (CTR2 bit 1).

PING-PONG Mode

PING-PONG mode is only valid in TRANSMIT mode. T8 and T16 must be programmed in SINGLE-PASS mode (CTR0, bit 6; CTR2, bit 6), and PING-PONG mode must be programmed in CTR1 bits [3:2]. You can begin the operation by enabling either T8 or T16 (CTR0, D7 or CTR2, D7). For example, if T8 is enabled, T8_OUT is set to this initial value (CTR1, bit 1).

According to T8_OUT's level, TC8H or TC8L is loaded into T8. After the terminal count is reached, T8 is disabled, and T16 is enabled. T16_OUT then switches to its initial value (CTR1, bit 0), data from TC16H and TC16L is loaded, and T16 starts to count. After T16 reaches the terminal count, it stops, T8 is enabled again, repeating the entire cycle. Interrupts can be allowed when T8 or T16 reaches terminal control (CTR0, bit 1; CTR2, bit 1). To stop the PING-PONG operation, write 00 to bits CTR1 bits [3:2]. See Figure 38 on page 112.

Enabling PING-PONG operation while the counter/timers are running may cause intermittent counter/timer function. Disable the counter/timers and reset the status flags before instituting this operation. **Note:**

Figure 38. PING-PONG Mode Diagram

Initiating PING-PONG Mode

First, ensure that both counter/timers are not running.

Follow the steps below to initiate the PING-PONG mode:

- 1. Set T8 into SINGLE-PASS mode (CTR0, bit 6)
- 2. Set T16 into SINGLE-PASS mode (CTR2, bit 6)
- 3. Set the PING-PONG mode (CTR1 bits [3:2])

These instructions are not consecutive and can occur in random order.

4. Finally, start PING-PONG mode by enabling either T8 (CTR0, D7) or T16 (CTR2, D7). The initial value of T8 or T16 must not be 1.

If you stop the timer and restart the timer, reload the initial value to avoid an unknown previous value.

During PING-PONG Mode

The enable bits of T8 and T16 (CTR0, D7; CTR2, D7) are set and cleared alternately by hardware. The time-out bits (CTR0, bit 5; CTR2, bit 5) are set every time the counter/ timers reach the terminal count.

Timer Output

The output logic for the timers is displayed in Figure 39 on page 113. P34 is used to output T8 OUT when bit 0 of CTR0 is set. P35 is used to output the value of T16 OUT when bit

0 of CTR2 is set. When bit 6 of CTR1 is set, P36 outputs the logic combination of T8_OUT and T16_OUT via bits [4:5] of CTR1.

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Counter/Timer Registers

The following sections describe each of the Timer/Counter registers in detail.

Timer 8 Capture High Register

The Timer 8 Capture High register (see Table 48) holds the captured data from the output of the 8-bit Counter/Timer 0. This register contains the number of counts when the input signal is 1.

This register is not reset after a Stop Mode Recovery. **Note:**

Table 48. Timer 8 Capture High Register (HI8)

Bit		6			З	כי				
Field	T8_Capture_HI									
Reset										
R/W	R	R	R	R	R	R	R	R		
Address	Bank D: 0Bh; Linear: D0Bh									
-----		\mathbf{r} .								

[7:0] 0hh–FFh **T8_Capture_HI**—Reads return captured data. Writes have no effect.

Timer 8 Capture Low Register

The Timer 8 Capture Low register (see Table 49 on page 115) holds the captured data from the output of the 8-bit Counter/Timer 0. Typically, this register contains the number of counts when the input signal is 0.

Table 49. Timer 8 Capture Low Register (LO8)

Timer 16 Capture High Register

The Timer 16 Capture High register (see Table 50) holds the captured data from the output of the 16-bit Counter/Timer 16. This register contains the most significant byte (MSB) of the data.

This register is not reset after a Stop Mode Recovery. **Note:**

Table 50. Timer 16 Capture High Register (HI16)

[7:0] 0hh–FFh **T16_Capture_HI**—Read returns captured data. Writes have no effect.

▶

Timer 16 Capture Low Register

The Timer 16 Capture Low register (see Table 51) holds the captured data from the output of the 16-bit Counter/Timer 16. This register contains the least significant byte (LSB) of the data.

This register is not reset after a Stop Mode Recovery. **Note:**

Counter/Timer 16 High Hold Register

The Counter/Timer 16 High Hold register (see Table 52) contains the high byte of the value loaded into the T16 timer.

This register is not reset after a Stop Mode Recovery. **Note:**

Table 52. Counter/Timer 16 High Hold Register (TC16H)

Counter/Timer 16 Low Hold Register

The Counter/Timer 16 Low Hold register (see Table 53) contains the low byte of the value loaded into the T16 timer.

 \blacktriangleright

This register is not reset after a Stop Mode Recovery. **Note:**

Table 53. Counter/Timer 16 Low Hold Register (TC16L)

Counter/Timer 8 High Hold Register

The Counter/Timer 8 High Hold register (see Table 54) contains the value to be counted while the T8 output is 1.

This register is not reset after a Stop Mode Recovery. **Note:**

Table 54. Counter/Timer 8 High Hold Register (TC8H)

Bit Position Value Description

[7:0] 0hh–FFh **T8_Level_HI**—Read/Write Data.

[▶]

Counter/Timer 8 Low Hold Register

The Counter/Timer 8 Low Hold register (see Table 55) contains the value to be counted while the T8 output is 0.

 \blacktriangleright

This register is not reset after a Stop Mode Recovery. **Note:**

Table 55. Counter/Timer 8 Low Hold Register (TC8L)

Counter/Timer 8 Control Register

The Counter/Timer 8 Control register (see Table 56) controls the timer function of the T8 timer.

Writing 1 to CTR0[5] is the only way to reset the Terminal Count status condition. Reset **Caution:** *this bit before using/enabling the counter/timers.*

You must be careful when using the OR or AND commands to manipulate CTR0, bit 5 and CTR1, bits 0 and 1 (DEMODULATION Mode). These instructions use a Read-Modify-Write sequence in which the current status from the CTR0 and CTR1 registers is ORed or ANDed with the designated value and then written back into the registers. **Note:**

Example: When the status of bit 5 is 1, a timer reset condition occurs.

Bit 7 6 5 4 3 2 1 0 Field T8 Enable SINGLE-PASS/ MODULO-N Time_Out | T8 _Clock | Capture_INT_Mask | Counter_INT_Mask | P34_Out **Reset** 000 0 0 0 0 0 **R/W** R/W R/W R/W R/W R/W R/W R/W R/W **Address** Bank D: 00h; Linear: D00h

Table 56. Counter/Timer 8 Control Register (CTR0)

T8 and T16 Common Functions Register

The T8 and T16 Common Functions register (CTR1) controls the functions in common with Timer 8 and Timer 16. Table 57 describes the bits for this register.

Note:

Be careful to differentiate TRANSMIT mode from DEMODULATION mode, as set by CTR1[7]. The functions of CTR1[6:0] and CTR2[6] are different depending on which mode is selected. Do not change from one mode to another without first disabling the counter/timers.

[▶]

Counter/Timer 16 Control Register

Table 58 describes the bits for the Counter/Timer 16 Control register (CTR2).

Table 58. Counter/Timer 16 Control Register (CTR2)

Timer 8/Timer 16 Control Register

The Timer 8/Timer 16 Counter/Timer register allows the start time of the T8 and T16 counters to be synchronized by simultaneously programming bits 7 and 6 to '1'. It also can freeze the T16 output value and change T8 DEMODULATION mode to capture one cycle of a carrier. Table 59 briefly describes the bits for this Bank D register.

Table 59. Timer 8/Timer 16 Control Register (CTR3)

Interrupts

The ZLF645 MCU features six interrupts (see Table 61 on page 129). These interrupts are maskable and prioritized (see Figure 40 on page 128).

The six interrupt sources are divided as follows:

- **•** Three sources are claimed by Port 3 lines P33:P31
- **•** Two by the counter/timers (see Table 61)
- One for low-voltage detection

P32 and UART receiver share the same interrupt. Only one interrupt can be selected as a source. When the UART receiver is enabled, P32 is no longer used as an interrupt source. The UART transmit interrupt and UART baud rate interrupt use the same interrupt as the P33 interrupt. The user selects which source triggers the interrupt. When bit 7 of UCTL is 1, the UART transmit interrupt is the source. When bit 7 of UCTL is 0 and bit 5 of UCTL is 1, the BRG interrupt is selected. The Interrupt Mask Register (globally or individually) enables or disables the six interrupt requests. The source for IRQ1 is determined by bit 1 of the Port 3 Mode register (P3M) and bit 4 of the SMR4 register.

If P3M[1]=0 (DIGITAL mode) and SMR4[4]=0, pin P33 is the IRQ1 source.

If P3M[1]=1 (ANALOG mode) or SMR4[4]=1 (SMR interrupt enabled), the output of the Stop Mode Recovery source logic is used as the source for the interrupt. For more details, see Stop Mode Recovery Interrupt on page 144.

Table 60. Interrupt Control Registers

Figure 40. Interrupt Block Diagram

Table 61. Interrupt Types, Sources, and Vectors

When more than one interrupt is pending, priorities are resolved by a programmable priority encoder controlled by the Interrupt Priority Register. An interrupt machine cycle activates when an interrupt request is granted. As a result, all subsequent interrupts are disabled, and the Program Counter and Status Flags are saved. The cycle then branches to the Program Memory vector location reserved for that interrupt.

All ZLF645 MCU interrupts are vectored through locations in the Program Memory. This memory location and the next byte contain the 16-bit address of the interrupt service routine for that particular interrupt request. To accommodate polled interrupt systems, interrupt inputs are masked, and the Interrupt Request Register is polled to determine which of the interrupt requests require service.

An interrupt resulting from AN1 is mapped into IRQ2, and an interrupt from AN2 is mapped into IRQ0. Interrupts IRQ2 and IRQ0 can be rising, falling, or both edge triggered. These interrupts are user-programmable. The software can poll to identify the state of the pin.

Programming bits for the Interrupt Edge Select are located in the IRQ register (R250), bit 6 and bit 7. Table 62 provides the configuration.

Table 62. Interrupt Request Register

Interrupt Priority Register

The Interrupt Priority register (see Table 63) defines which interrupt holds the highest priority. Interrupts are divided into three groups of two—Group A, Group B, and Group C.

IPR bits 4, 3, and 0 determine which interrupt group has priority. For example, if interrupts IRQ5, IRQ1, and IRQ0 occur simultaneously when IPR[4:3,0]=001b, the interrupts are serviced in the following order: IRQ1, IRQ0, IRQ5.

IPR bits 5, 2, and 1 determine which interrupt within each group has higher priority.

Table 63. Interrupt Priority Register (IPR)

Interrupt Request Register

Bit 7 and Bit 6 of the Interrupt Request register (see Table 64) are used to configure the edge detection of the interrupts for Port 3, bit 1 and Port 3, bit 2. The remaining bits (5 through 0) indicate the status of the interrupt. When an interrupt is serviced, the hardware automatically clears the bit to 0. Writing 1 to any of these bits generates an interrupt if the appropriate bits in the Interrupt Mask register are enabled. Writing 0 to these bits clears the interrupts.

Bit	7	6	5	4	3	$\mathbf{2}$		$\bf{0}$		
Field	Interrupt Edge		IRQ5	IRQ4	IRQ3	IRQ ₂	IRQ1	IRQ0		
Reset	0	0	Ω	0	$\mathbf 0$	0	$\mathbf 0$	0		
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
Address	Bank Independent: FAh; Linear: 0FAh									
		Description								
Bit Position	Value									

Table 64. Interrupt Request Register (IRQ)

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The IRQ register is protected from change until an EI instruction is executed once. **Note:**

Interrupt Mask Register

Bits [5:0] are used to enable the interrupt. Bit 7 is the status of the master interrupt. When reset, all interrupts are disabled. When writing 1 to bit 7, you must also execute the EI instruction to enable interrupts (see Table 65).

Table 65. Interrupt Mask Register (IMR)

Clock

ZLF645 MCUs on-chip oscillator has a high-gain, parallel-resonant amplifier for connecting to a crystal, ceramic resonator, or any suitable external clock source $(XTAL1 = Input, XTAL2 = Output).$

Crystal Specification

The crystal must be AT cut, 1 MHz to 8 MHz (maximum), with a series resistance (RS) less than or equal to 100 Ω . The on-chip oscillator can be driven with a suitable external clock source.

The crystal must be connected across XTAL1 and XTAL2 pins using the recommended capacitors from each pin to ground. The typical capacitor value is 10 pF for 8 MHz. Check with the crystal supplier for the optimum capacitance.

*Note: preliminary value, including pin parasitics.

Figure 41. Oscillator Configuration

Maxim's IR MCU supports crystal, resonator, and oscillator. Most resonators have a frequency tolerance of less than ±0.5%, which is enough for a remote control application. Resonator has a very fast startup time, which is around few hundred microseconds. Most crystals have a frequency tolerance of less than 50 ppm $(\pm 0.005\%)$. However, crystal needs longer startup time than the resonator. The large loading capacitance slows down the oscillation startup time. Clock oscillation must be stable before the CPU begins instruction execution. If oscillation is not present or not stable before the chip completes timeout of its Power-On Reset (POR) period, the chip's behavior could be indeterminate.

Maxim*®* recommends not to use more than 10 pF loading capacitor for the crystal. If the stray capacitance of the PCB or the crystal is high, the loading capacitance C1 and C2 must be reduced further to ensure stable oscillation before the T_{POR} (POR time is typically 5-6 ms. For more details, see Table 81 on page 169.).

For Stop Mode Recovery operation, bit 5 of SMR register allows you to select the Stop Mode Recovery delay, which is the T_{POR} . If Stop Mode Recovery delay is not selected, the MCU executes instruction immediately after it wakes up from the STOP mode. If resonator or crystal is used as a clock source then STOP mode recovery delay needs to be selected (Bit 5 of $SMR = 1$).

For both resonator and crystal oscillator, the oscillation ground must go directly to the ground pin of the microcontroller. The oscillation ground must use the shortest distance from the microcontroller ground pin and it must be isolated from other connections.

Crystal 1 Oscillator Pin (XTAL1)

The Crystal 1 Oscillator time-based input pin connects a parallel-resonant crystal or ceramic resonator to the on-chip oscillator input. Additionally, an optional external singlephase clock can be connected to the on-chip oscillator input.

Crystal 2 Oscillator Pin (XTAL2)

The Crystal 2 Oscillator time-based output pin connects a parallel-resonant crystal or ceramic resonant to the on-chip oscillator output.

Internal Clock Signals (SCLK and TCLK)

The CPU and internal peripherals are driven by the internal SCLK signal during normal execution. During HALT mode, the interrupt logic is driven by the internal TCLK signal. The frequency of these signals with respect to the XTAL1 clock input is selectable either by programming bit 2 of the Flash's User Option Byte 1 for no division of the XTAL1 signal input, dividing it by a factor of two, and optionally by applying an additional divide-by-16 prescaler enabled through SMR register bit 0 (see Table 69 on page 146), as displayed in Figure 42. Selecting the divide-by-16 prescaler reduces device power drawduring normal operation and HALT mode. The prescaler is disabled by a POR or Stop Mode Recovery.

Figure 42. SCLK/TCLK Circuit

Reset and Power Management

The ZLF645 MCU provides the following reduced-power modes, power monitoring, and reset features:

- **• Voltage Brownout Standby**—Stops the oscillator and internal clock when the power level drops below the VBO low voltage detect point. Initiates a power-on reset when power is restored above the VBO detect point.
- **• STOP Mode**—Stops the clock and oscillator, reduces the MCU supply current to a very low level until a power-on reset or Stop Mode Recovery occurs.
- **• HALT Mode**—Stops the internal clock to the CPU until an enabled interrupt request is received.
- **• Voltage Detection**—Optionally sets a flag if a low- or high-voltage condition occurs. The low-voltage detection flag can generate an interrupt request, if enabled.
- **Power-On Reset—Starts the oscillator and internal clock, and initializes the system to** its power-on reset defaults.
- **• Watchdog Timer**—Optionally generates a Power-On Reset if the program fails to execute the WDT instruction within a specified time interval.
- **• Stop Mode Recovery**—Restarts the oscillator and internal clock, and initializes most of the system to its power-on reset defaults. Some register values are not reset by a Stop Mode Recovery.

For supply current values under various conditions, see DC Characteristics *on page 165.* **Note:**

> Figure 43 on page 138 displays the Power-On Reset sources. Table 66 lists control registers for reset and power management features. Some features are affected by registers described in other chapters.

Table 66. Reset and Power Management Registers

Table 66. Reset and Power Management Registers (Continued)

***CLR1 and CLR2 enable the WDT/POR and 18 Clock Reset timers, respectively, on a Low-to-High input transition. Figure 43. Resets and Watchdog Timer**

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Voltage Brownout Standby

An on-chip voltage comparator circuit (VBO) checks that the V_{DD} is at the required level for correct operation of the device in terms of Flash memory reads. A second on-chip comparator circuit (subVBO) checks that the V_{DD} level is high enough for proper operation of the VBO circuit. If the V_{DD} level drops below the VBO trip point, the ZLF645 will be held in a reset state as long as V_{DD} remains below this trip point value, and the XTAL1 and XTAL2 oscillator circuitry will be disabled thereby stopping the clock input to the ZLF645 and saving power. If the V_{DD} level continues to drop below the subVBO trip point, the ZLF645 will remain in a reset state and the VBO comparator circuit will be disabled for further power savings. When the power level returns to a value above the VBO trip point, the device performs a power-on reset and functions normally.

STOP Mode

STOP instruction turns OFF the internal clock and external crystal oscillation, thus reducing the MCU supply current to a very low level. For STOP mode current specifications, see Table 80 on page 165.

To enter STOP mode, first flush the instruction pipeline to avoid suspending execution in mid-instruction. Execute a NOP instruction (OpCode $=$ FFh) immediately before the appropriate sleep instruction, as given below:

STOP mode is terminated only by a reset, such as WDT time-out, POR, or one of the Stop Mode Recovery events as described in Stop Mode Recovery Event Sources on page 144. This condition causes the processor to restart the application program at address 000Ch.

Unlike a normal POR or WDT reset, a Stop Mode Recovery reset does not reset the contents of some registers and bits. Register bits not reset by a Stop Mode Recovery are highlighted in grey in the register tables. Register bit SMR[7] is set to 1 by a Stop Mode Recovery.

HALT Mode

HALT instruction turns off the internal CPU clock, but not the XTAL oscillation. The counter/timers, UART, and interrupts (IRQ0, IRQ1, IRQ2, IRQ3, IRQ4, and IRQ5) remain active. The devices are recovered by interrupts, either externally or internally generated. An interrupt request must be executed (enabled) to exit HALT mode. After the interrupt service routine, the program continues from the instruction after HALT mode.

To enter HALT mode, first flush the instruction pipeline to avoid suspending execution in mid-instruction. Execute a NOP instruction (OpCode $=$ FFh) immediately before the appropriate sleep instruction, as given below:

FF NOP ; clear the pipeline 7F **HALT HALT** *i* enter HALT mode

Power consumption during HALT mode can be reduced by first setting SMR[0]=1 to enable the divide-by-16 clock prescaler.

Voltage Detection

The Low-Voltage Detection register (LVD, register 0Ch at the expanded register bank 0Dh) provides an option to monitor the V_{DD} voltage. The voltage detection is enabled when bit 0 of LVD register is set. After voltage detection is enabled, the V_{DD} level is monitored in real time. The HVD flag (bit 2 of the LVD register) is set only if V_{DD} is higher than V_{HVD} . The LVD flag (bit 1 of the LVD register) is set only if V_{DD} is lower than the V_{LVD} . When voltage detection is enabled, the LVD flag also triggers IRQ5. The IRQ bit 5 latches the low voltage condition until it is cleared by instructions or reset. The IRQ5 interrupt is served if it is enabled in the IMR register. Otherwise, bit 5 of IRQ register is latched as a flag only.

Do not modify register P01M while checking a low voltage condition. Switching noise from Port 0 can trigger the LVD flag. **Note:**

Bit		6	5	4	3	2		0	
Field	Reserved					High-Battery Detect	Low-Battery Detect	Voltage Detect Enable	
Reset							0		
R/W	R	R	R	R	R	R	R	R/W	
Address	Bank D: 0Ch; Linear: D0Ch								

Table 67. Low-Voltage Detection Register (LVD)

Power-On Reset Timer

When power is initially applied to the device, a timer circuit clocked by a dedicated on-board RC-oscillator provides the POR timer function. The POR timer circuit is a oneshot timer that keeps the internal reset signal asserted long enough for V_{DD} and the oscillator circuit to stabilize before instruction execution begins.

The reset timer is triggered by one the following conditions:

- **•** Initial power-on or recovery from a VBO/standby condition.
- Stop Mode Recovery (if register bit $SMR[5] = 1$).
- **•** Watchdog Timer time-out.

SMR[5] can be cleared to 0 to bypass the POR timer on a Stop Mode Recovery. This must only be done when using an external clock that does not require a startup delay.

Failure of an application to provide a stable oscillating clock input to XTAL1 before the **Caution:***end of the ZLF645's POR period may result in an indeterminate chip behavior and must be avoided. For details on the POR timing range, see* Table 81 *on page 169 in the* Electrical Characteristics *chapter.*

Watchdog Timer

The Watchdog Timer (WDT) is a retriggerable one-shot timer that resets the Z8 LXMC CPU if it reaches its terminal count. The WDT must initially be enabled by executing the WDT instruction. On subsequent executions of the WDT instruction, the WDT is refreshed. The WDT circuit is driven by an on-board RC-oscillator. The WDT instruction affects the Zero (Z) , Sign (S) , and Overflow (V) flags.

The POR clock source is an internal RC-oscillator. Bits 0 and 1 of the WDT register control a tap circuit that determines the minimum time-out period. Bit 2 determines whether the WDT is active during HALT, and bit 3 determines WDT activity during STOP mode. Bits 4 through 7 are reserved (see Table 68 on page 142). This register is accessible only during the first 60 processor cycles (120 XTAL clocks) from the execution of the first instruction after power-on reset, watchdog timer Reset, or a Stop Mode Recovery (see Fast Stop Mode Recovery). After this point, the register cannot be modified by any means. The

WDTMR register cannot be read. The register is located in Bank F of the Expanded Register Group at address location 0Fh.

This register is not reset after a Stop Mode Recovery. **Note:**

Table 68. Watchdog Timer Mode Register (WDTMR)

$[2]$		WDT During HALT Mode —Determines if WDT is active during HALT mode. See Figure 43 on page 138.
	0	Off.
	1	WDT active during HALT mode.
[6:4], [1:0]		Time-Out Select—Selects the WDT time period (see Note below).
		000_00 5 ms minimum
		000 01 10 ms minimum
		000 10 20 ms minimum
		000 11 80 ms minimum
		001 XX 320 ms minimum
		010 XX 1, 280 ms minimum
		100 XX 5,120 ms minimum

Although not explicitly shown above, if any two bits of bits 6 through 4 are programmed to 1 or if all three bits are programmed to 0, then the time-out period depends on bits 1 and 0 only as shown for the [6:4]=000 case. **Note:**

Using the Watchdog Timer As a Stop Mode Recovery Source

As mentioned previously, timeout of the Watchdog Timer generates a chip reset that removes the ZLF645 from STOP mode. This feature is used to configure the ZLF645 to automatically exit STOP mode, once it is in STOP mode. This is done within a set maximum period of time based upon the Watchdog Timer timeout setting. In this way, the

ZLF645 is configured to periodically enter and exit STOP mode until some action is necessary by the application. Follow the steps below to configure the ZLF645 for this mode of operation:

- 1. Provide program code that, within 60 processor clock cycles of the start of code execution after reset, programs bits 6 through bit 4 and bit 1 through 0 of the Watchdog Timer Register (WDTMR) with the time-out setting required. Also ensure bit 3 of the WDTMR register is '1', configuring the WDT for counting once enabled, even with the ZLF645 in STOP mode.
- 2. Execute the WDT instruction to enable counting of the Watchdog Timer.
- 3. Include a STOP instruction after the program code for steps 1 & 2 above, that puts the ZLF645 into STOP mode when no action is required by the ZLF645.

Once the Watchdog Timer has been enabled through the WDT instruction it will begin counting and continue counting even after the ZLF645 has entered STOP mode, due to the bit 3 of the WDTMR register being '1'. Once the WDT has reached its time-out value, it will initiate a reset condition that takes the ZLF645 out of STOP mode. This reset condition will not cause the WDTMR register to be reset and it will retain its previous programming. Upon completion of the reset the WDT will enter a disabled state and stop counting. Execution of a new WDT instruction is necessary to cause the Watchdog Timer to reset to its start count state and to start counting again. It is important to note that if a time-out of the Watchdog Timer occurs with STOP mode inactive, the reset generated will cause a reset of the WDTMR register and it will not retain its previously programmed value. In either case of STOP mode being active or inactive during Watchdog Timer time-out, the WDT will go to a disabled state upon completion of reset.

Reset/Stop Mode Recovery Status

Read-only bit SMR[7]=0, if the previous reset was initiated by a Power-On Reset (including Voltage Brownout or WDT resets). $SMR[7]=1$, if the previous reset was initiated by a Stop Mode Recovery. A power-on, Voltage Brownout, or WDT reset restores all registers to their Power-On Reset defaults. A Stop Mode Recovery restores most registers to their Power-On Reset defaults. Register bits not reset by a Stop Mode Recovery are highlighted in grey in the register tables. Register bit SMR[7] is set to 1 instead of reset by a Stop Mode Recovery.

Fast Stop Mode Recovery

SMR[5] can be cleared to 0 before entering STOP mode to bypass the default T_{POR} reset timer on Stop Mode Recovery. See Voltage Brownout Standby on page 139.

If SMR[5]=0, the Stop Mode Recovery source must be kept active for at least 10 input clock periods (TpC).

SMR[5] must be set to 1 if using a crystal or resonator clock source. The T_{POR} delay *allows the clock source to stabilize before executing instructions.* **Note:**

Stop Mode Recovery Interrupt

Software can set register bit $SMR4[4] = 1$ to enable routing of Stop Mode Recovery events to IRQ1 and to Port 3, Pin 3. In this configuration, if an IRQ1 interrupt occurs, register bit $P3[3] = 0$ indicates that a Stop Mode Recovery event is occurring.

Stop Mode Recovery Event Sources

Any Port 2 or Port 3 input pin can be configured to generate a Stop Mode Recovery event, either individually or in various logical combinations. The ZLF645 MCU provides the following registers for Stop Mode Recovery source configuration and status:

- **• SMR Register**—Selects one Port 3, Pin 1–3 pin state or one of three Port 2 pin logical combinations to generate an event when a defined 0 or 1 level occurs.
- **SMR1 Register**—Configures one or more Port 2 input pins (0–7) to latch the latest read or write value and generate an event when the pin state changes.
- **• SMR2 Register**—Selects one of seven Port 2 and 3 pin logical combinations to generate an event when a defined 0 or 1 level occurs.
- **SMR3 Register—Configures one or more Port 3 input pins (0–3) to latch the latest** read or write value and generates an event when the pin state changes.
- **SMR4 Register—Enables routing of SMR events to IRQ1. Indicates whether port data** has been latched for SMR1 or SMR3 event monitoring, and whether the latch was on a port read or write.

A Stop Mode Recovery event occurs if any of the sources defined in the SMR, SMR1, SMR2, and SMR3 registers are active.

SMR Register Events

The SMR register function is similar to the standard Stop Mode Recovery feature used in previous Z8® CPU-compatible parts. Register bits SMR[4:2] are set to select one of six event modes, as displayed in Figure 44 on page 145. The output of the corresponding logic is compared to the state of SMR[6]; when they are the same, a Stop Mode Recovery event is generated. If SMR[4:2]=000, no event source is selected by SMR.

The state SMR[4:2]=001 is reserved and selects no event in this device. The logic configured by the SMR register ignores any port pins that are configured as output or selected as source pins in registers SMR1 or SMR3. The SMR register is summarized in Table 69 on page 146.

[**Table 69. Stop Mode Recovery Register (SMR)**

SMR1 Register Events

The SMR1 register can be used to configure one or more Port 2 pins to be compared with a written or sampled reference value and generate a Stop Mode Recovery event when the pin state differs from the reference value.

To configure a Port 2 pin as an SMR1 event source, ensure it is configured as an input in the P2M register, then set the corresponding SMR1 register bit. By default, a Stop Mode Recovery event occurs when the pin's state is zero.

After a Port 2 pin is configured as an SMR1 source, any subsequent read from or write to the P2 register latches the read or write value for reference. A Stop Mode Recovery event occurs when the pin's state differs from the last reference value latched. The SMR1 source logic is displayed in Figure 45 on page 149.

The program can read register bits SMR4[1:0] to determine whether the Port 2 pins trigger a Stop Mode Recovery on a change from the last read value $(SMR4[1:0]=01)$, or on a change from the last written value (SMR4 $[1:0]$ =10). Software can clear SMR4 $[1:0]$ to 00 to restore the default behavior (configured pins trigger when their state is 0). The SMR1 register is summarized in Table 70 on page 150.

After the following example code is executed, a 1 on P20 will wake the part from STOP mode:

LD P2M, #%FF ;Set Port 2 to inputs. SRP #%0F ;Point to expanded bank F LD SMR1, #%01 ;Select P20 for SMR1. SRP #%00 ;Point to bank 0 LD P2, #%00 ;Write 00h to Port 2, so the P20 reference ;value is 0, and a 1 on P20 wakes the part. NOP STOP

After the following example code is executed when the value of P2 is 00h, a 1 on P20 will wake the part from STOP mode:

```
LD P2M, #%FF ;Set ports to inputs.
SRP #%0F : Point to expanded bank F
LD SMR1, #%01 ; Select P20 for SMR1.
SRP #%00 ;Point to bank 0
LD R6, P2 : If a 0 is read from Port 2, the P20 reference
              ;value is 0, so a 1 on P20 wakes the part.
NOP
STOP
```


Individual Port 2 Pin SMR Logic, $n = 0 - 7$

Figure 45. SMR1 Register-Controlled Event Sources

Table 70. Stop Mode Recovery Register 1 (SMR1)

This register is not reset after a Stop Mode Recovery. **Note:**

SMR2 Register Events

The SMR2 register function is similar to the standard Stop Mode Recovery feature used in previous Z8 CPU-compatible parts. Register bits SMR2[4:2] are set to select one of seven event modes, as displayed in Figure 46. The output of the corresponding logic is compared to the state of SMR2[6]; when they are the same, a Stop Mode Recovery event is generated. If SMR2[4:2]=000, no event source is selected by SMR2.

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The logic configured by the SMR2 register ignores any port pins that are configured as an output, or that are selected as source pins in registers SMR1 or SMR3. The SMR2 register is summarized in Table 71 on page 152.

Figure 46. SMR2 Register-Controlled Event Sources

Table 71. Stop Mode Recovery Register 2 (SMR2)

Bit	$\overline{7}$	66	5	$\overline{\mathbf{4}}$	$\mathbf{3}$	$\overline{2}$	1	$\bf{0}$		
Field	Reserved	Stop Mode Recovery Level 2	Reserved	Stop Mode Recovery Source			Reserved			
Reset	X	0	X	0	0	0	X	X		
R/W		W		W	W	W				
Address		Bank F: 0Dh; Linear: F0Dh								
Bit Position	Value	Description								
$[7]$		Reserved-Read is undefined; write must be 0.								
[6]	0 1	Stop Mode Recovery Level 2 Selects whether an SMR2[4:2]-selected SMR is initiated by a Low or High level at the XOR-gate input (see Figure 46 on page 151). Low. High.								
$[5]$		Reserved-Read is undefined; Must be written to 1.								
[4:2]	000 001 010 011 100 101 110 111	Stop Mode Recovery Source Specifies a Stop Mode Recovery wake-up source at the XOR gate input (see Figure 46 on page 151). Additional sources can be selected by SMR, SMR1, and SMR3 registers. If more than one source is selected, any selected source event causes a Stop Mode Recovery. The following equations ignore any Port pin that is selected in register SMR1 or configured as an output. No SMR2 register source selected. NAND of P23:P20. NAND of P27:P20. NOR of P33:P31. NAND of P33:P31. NOR of P33:P31, P00, P07. NAND of P33:P31, P00, P07. NAND of P33:P31, P22:P20.								
[1:0]		Reserved-Read is undefined; write must be 00b.								

 \blacktriangleright

This register is not reset after a Stop Mode Recovery. **Note:**

SMR3 Register Events

The SMR3 register can be used to configure one or more of Port 3, pins 0–3 to be compared to a written or sampled reference value and generate a Stop Mode Recovery event when the pin state differs from the reference value.

To configure a Port 3 input pin as an SMR3 event source set the corresponding SMR3 register bit. By default, a Stop Mode Recovery event occurs when the pin's state is zero.

After a Port 3 pin is configured as an SMR3 source, any subsequent read from or write to the P2 register latches the read or written value for reference. A Stop Mode Recovery event occurs when the pin's state differs from the last reference value latched. The SMR3 source logic is displayed in Figure 47.

The program can read register bits SMR4[3:2] to determine whether the Port 3 pins trigger a Stop Mode Recovery on a change from the last read value $(SMR4[3:2]=01)$, or on a change from the last written value (SMR4[3:2]=10). Software can clear SMR4[3:2] to 00 to restore the default behavior (configured pins trigger when their state is 0).

The SMR3 register is summarized in Table 69 on page 146.

After the following example code is executed, a 1 on P30 will wake the part from STOP mode.

```
LD SMR3, #%01 ;Select P30 from SMR3.
               Write 00h to Port 3, so the P30 reference
                ;value is 0, and a 1 on P30 wakes the part.
NOP
STOP
```
After the following example code is executed when the value of P3 is 00h, a 1 on P30 will wake the part from STOP mode.

```
LD SMR3, #%01 ; Select P30 for SMR3.
LD R6, P3 : If a 0 is read from Port 3, the P30 reference
               ;value is 0, so a 1 on P30 wakes the part.
NOP
STOP
```


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Individual Port 3 Pin SMR Logic, $n = 0 - 3$

Figure 47. SMR3 Register-Controlled Event Sources

Table 72. Stop Mode Recovery Register 3 (SMR3)

 \blacktriangleright

This register is not reset after a Stop Mode Recovery. **Note:**

Stop Mode Recovery Register 4

The Stop Mode Recovery register 4 (see Table 73) enables the SMR interrupt source and indicates the reference value status for registers SMR1 and SMR3.

Table 73. Stop Mode Recovery Register 4 (SMR4)

This register is not reset after a Stop Mode Recovery. **Note:**

Z8 LXMC CPU Programming Summary

The following sections provide a summary of information useful for programming the Z8 LXMC CPU included in this device. For more details on the Z8 LXMC CPU and its instruction set, refer to *Z8*® *LXMC CPU Core User Manual (UM0215).*

Addressing Notation

Table 74 summarizes Z8 LXMC CPU addressing modes and symbolic notation. The text variable *n* represents a decimal number; *aa* represents a hexadecimal address; and *LABEL* represents a label defined elsewhere in the assembly source.

In reference notation, *only* lowercase is used to distinguish 4-bit addressed working registers (r1, r2) from 8-bit addressed registers (R1, R2). The numerals 1 and 2, respectively, indicate whether the register is used for destination or source addressing.

Table 74. Symbolic Notation for Operands (Continued)

Table 74. Symbolic Notation for Operands (Continued)

Table 75 lists additional symbols that are used throughout the instruction set summary.

Table 75. Additional Symbols

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Table 75. Additional Symbols (Continued)

Symbol	Definition
	Exchange of two values
$\tilde{}$	One's complement unary operator

Flags Register

The Flags register (see Table 76) informs the current status of the Z8[®] CPU. It contains six bits of status information.

Table 76. Flags Register (FLAGS)

Bit	$\overline{7}$	6	5	4	3	$\overline{2}$	$\mathbf 1$	$\bf{0}$	
Field	C	Ζ	S	Ω	D	H	F ₁	F ₂	
Reset	X	X	X	X	X	X	X	X	
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Address	Bank Independent: FCh; Linear: 0FCh								
Bit Position	Value Description								
$[7]$	0 1	Carry Flag (C) Set when the result of an arithmetic operation generates a carry out of or a borrow into the high-order bit (bit 7) of the result. Also used in rotate and shift instructions. Flag Clear Flag Set							
[6]	$\mathbf 0$ 1	Zero Flag (Z) Set when the result of an arithmetic operation is 0. Flag Clear Flag Set							
[5]	$\mathbf 0$ 1	Sign Flag (S) Stores the value of the most significant bit (msb) following an arithmetic, logical, rotate, or shift instruction. Flag Clear Flag Set							
$[4]$	0 1	Overflow Flag (O) Set when the result of an arithmetic operation is greater than 127. Flag Clear Flag Set							

Condition Codes

The C, Z, S, and V flags control the operation of the conditional jump (JP cc and JR cc) instructions. Sixteen frequently useful functions of the flag settings are encoded in a 4-bit field called the condition code (cc). Table 77 summarizes the condition codes. Some binary condition codes can be created using more than one assembly code mnemonic. The result of the flag test operation determines if the conditional jump executes.

Table 77. Condition Codes

Table 77. Condition Codes (Continued)

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Electrical Characteristics

Absolute Maximum Ratings

A stress greater than listed in Table 78 may cause permanent damage to the device. Functional operation of the device at any condition outside those indicated in the operational sections of these specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. For improved reliability, unused inputs must be tied to one of the supply voltages (V_{DD} or V_{SS}).

Table 78. Absolute Maximum Ratings

Standard Test Conditions

The characteristics listed in this product specification apply for standard test conditions. All voltages are referenced to Ground. Positive current flows into the referenced pin (see Figure 48).

Figure 48. Test Load Diagram

Capacitance

Table 79 lists the capacitance.

Table 79. Capacitance

and P33.

DC Characteristics

Table 80 describes the direct current (DC) characteristics of the ZLF645 Flash MCU.

Table 80. DC Characteristics

Table 80. DC Characteristics (Continued)

Table 80. DC Characteristics (Continued)

Notes

1. Maxim[®] recommends adding a filter capacitor (minimum 0.1 μ F), physically close to V_{DD} and V_{SS} if operating voltage fluctuations are anticipated, such as those resulting from driving an infrared LED.

2. All outputs unloaded, inputs at rail.

3. CL1 = CL2 = 100 pF.

4. Oscillator stopped.

5. Oscillator stops when V_{DD} falls below V_{BO} limit.

6. For reference, under typical process, 25 °C temperature, and a voltage of 2.8 V, the voltage that would be seen at P31 under an input current of 10 uA or 100 uA would be ~480 mV or .560 mV, respectively.

7. 8.0-MHz XTAL1 input clock frequency with 4-MHz system clock

8. 8.0-MHz XTAL1 input clock frequency with 8-MHz system clock

AC Characteristics

Figure 49 and Table 81 lists the alternating current (AC) characteristics of ZLF645 Flash MCU.

Figure 49. AC Timing Diagram

Table 81. Clock, Reset, Timers, and SMR Timing

Table 81. Clock, Reset, Timers, and SMR Timing (Continued)

Notes

1. Timing Reference uses 0.9 V_{CC} for a logic 1 and 0.1 V_{CC} for a logic 0.

2. Interrupt request through Port 3 (P33:P31).

3. SMR – bit $5 = 1$.

4. $SMR - bit 5 = 0$.

5. If bit 1 of the SMR register is programmed to 1, this value is 2.5 ms as measured from the time the oscillator input to XTAL1 reaches a peak to peak voltage oscillation of at least 300 mV.

Table 82. Flash Memory Electrical Characteristics and Timing

Flash Option Bits

Programmable Flash Option Bits allow user configuration of certain aspects of ZLF645 MCU functionality. This configuration data is stored in the Flash memory Information Block and then read into option byte shadow registers during the last portion of the ZLF645 MCUs reset period.

Features available for control through the Flash Option Bits include:

- Port 0 low nibble pull-ups
- Port 0 high nibble pull-ups
- Port 1 low nibble pull-ups
- **•** Port 1 high nibble pull-ups
- **•** Port 2 pull-ups
- Port 3 low nibble pull-ups
- **•** Port 4 pull-ups
- **•** WDT always enabled
- Flash protect entire main memory
- **•** Flash protect lower half main memory
- **•** XTAL1 to System Clock (no division enable)
- **•** 16-bit Stack addressiblity enable

Operation

Option Bit Shadow Register Loading By Reset

For each Flash memory option bit, there is an associated option bit shadow register that is used to register the value of the option bit. The output of the option bit shadow registers are used by the ZLF645 MCU to enable various features and functions for the ZLF645 MCU. Each time the Flash Memory Information Block Option Bits are programmed or erased, the device must be reset for the change in ZLF645 configuration to take effect.

A POR or Stop Mode Recovery Reset with SMR bit 5 set to 1, loads the option bits from the Flash memory to the Option Bit Shadow registers during the last few clock cycles of the reset period. In some cases, in order to provide a required value before being loaded, the Option Bit Shadow registers are reset to a predefined value on the start of the reset period.

The Option Bit Shadow registers are part of the ZLF645's Register File and are accessible for read/write access.

User Option Bit Locations in Flash Memory

The user option bits are located in the upper two bytes of the Information block, address FFh and FEh.

User Option Bit Shadow Register Access

Except for bits 0 and 1 of the User Option Byte 1 Shadow register, the CPU has full read/write access to all the User Option Byte Shadow registers at the address locations given in register tables for each register.

User Option Byte 0 and Option Byte 0 Shadow Register Definitions

This option byte allows user control over the enabling of the ZLF645's I/O pull-ups and the conditions under which the devices watchdog timer is enabled. For its associated shadow registers, until the registers are loaded with their corresponding option bit values, their outputs will be in an unknown state.

Bit								
Field	WDT	P4PU	P ₃ PU	P ₂ PU	P1HPU	I P1LPU	P0HPU	POLPU
Erased State								
Flash Address	Flash Memory Information Area address: FEH							

Table 83. User Option Byte 0 (OPT0)

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P2PU = 1: Port 2 Pull-ups disabled. P2PU = 0: Port 2 Pull-ups enabled. P1HPU = 1: Port 1 high nibble Pull-ups disabled P1HPU = 0: Port 1 high nibble Pull-ups enabled. P1LPU = 1: Port 1 low nibble Pull-ups disabled. P1LPU = 0: Port 1 low nibble Pull-ups enabled. P0HPU = 1: Port 0 high nibble Pull-ups disabled. P0HPU = 0: Port 0 high nibble Pull-ups enabled. P0LPU = 1: Port 0 low nibble Pull-ups disabled. P0LPU = 0: Port 0 low nibble Pull-ups enabled.

User Option Byte 1 and Option Byte 1 Shadow Register Definitions

User Option byte allows the enabling of various features including protecting the Flash's main memory from read operations through either of the ZLF645's Flash access interfaces. For it's associated shadow registers, until the registers are loaded with their corresponding option bit values, their outputs will be in an unknown state.

Table 85 describes User Option byte 1 function.

This byte can be programmed and erased (by Page 3 erase) only through the ICP. **Note:**

> During device Power-on Reset, bit 1 and bit 0 value of this Flash Option byte are sampled into flip-flops, whose outputs control the Flash memory protect function. User codes can read the flip-flop values, by reading from a uniquely assigned peripheral register address.

User codes cannot over-write the flip-flop values to change this Flash memory protect function. **Note:**

Table 85. User Option Byte 1 (OPT1)

[0] 1 0 **FLRWP**—Flash Main Memory Protect Flash Main Memory and Information Area Page 3 can be read, programmed, and erased by both the Flash Byte Programming interface or through the ICP interface. Reads and writes to the Flash main memory and writes and erasures to Information Area Page 3 by the ICP or Flash Byte Programming interfaces is disabled unless, with this bit 0, a main memory mass erase is completed first. A main memory mass erase causes resetting of this bit value in the Option Byte 1 shadow register to a 1 but does not effect the corresponding Flash memory bit. Once the Option Byte 1 shadow register bit is reset, the ICP or Flash Byte Programming interface is allowed full read, write, and erase access to the Flash's main memory and to Page 3 of the Information Area and can reset the corresponding Flash memory bit.

Table 86. User Option Byte 1 Shadow Register (OPT1SR)

RESERVED bits when read by the CPU will return 0 and when written have no effect. **Note:**

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Packaging

Figure 50 displays the 20-pin quad flat no-lead (QFN) package for the ZLF645 Series of Flash MCUs.

1. CONTROLLING DIMENSIONS: mm 2 MAX COPLANARITY $\frac{0.08 \text{ mm}}{0.003}$

 $\ensuremath{\mathsf{INCH}}$

 $\textsf{NOM}{}$

0.073

 0.005

0.068

 0.012

 0.006

0.283

0.209

0.307

0.030

0.031

0.0256 BSC

MAX

 0.078

 0.008

 0.072

 0.015

 0.009

0.289

 0.212

 0.311

0.037

0.032

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Figure 51 displays the 20-pin shrink small outline package (SSOP) for the ZLF645 Series of Flash MCUs.

Figure 51. 20-Pin SSOP Package Diagram

CONTROLLING DIMENSIONS : MM
LEADS ARE COPLANAR WITHIN .004 INCH.

Figure 52. 20-Pin SOIC Package Diagram

Figure 53 displays the 20-pin plastic dual inline package (PDIP) for the ZLF645 Series of Flash MCUs.

Figure 53. 20-Pin PDIP Package Diagram

Figure 54 displays the 28-pin shrink small outline package (SSOP) for the ZLF645 Series of Flash MCUs.

SEATING PLANE

CONTROLLING DIMENSIONS: MM LEADS ARE COPLANAR WITHIN .004 INCHES.

DETAIL 'A'

 $0 - 8$

L

Figure 54. 28-Pin SSOP Package Diagram

Figure 55. 28-Pin SOIC Package Diagram

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Figure 56 displays the 28-pin plastic dual inline package (PDIP) for the ZLF645 Series of Flash MCUs.

Note: Zllog supplies both options for production. Component layout
PCB design should cover bigger option 01.

Figure 56. 28-Pin PDIP Package Diagram

Figure 57. 48-Pin SSOP Package Diagram

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Ordering Information

Table 87 lists the part numbers for ZLF645 Series of Flash MCUs and a brief description of each part.

Table 87. ZLF645 Flash MCU Part Numbers Description

Table 87. ZLF645 Flash MCU Part Numbers Description (Continued)

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Part Number Description

Maxim*®* part numbers consist of a number of components as shown below:

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